Observability Fast Robots, ECE4160/5160, MAE 4190/5190

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Class Action Items

- Lab 5 check in: how is everything going?
- I am hosting additional open hours tomorrow 8:30-11am and Sunday 6-8pm (moved from the original 11am-1pm), multiple requests for evening hours
- Tuesday's class: how are we feeling about probability?





Linear Systems — where are we?

- Linear systems review
- Eigenvectors and eigenvalues
- Stability
- Discrete time systems
- Linearizing nonlinear systems
- Controllability
- LQR control
- Observability

Based on "Control Bootcamp", Steve Brunton, UW https://www.youtube.com/watch?v=Pi7l8mMjYVE

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$$\dot{x} = Ax + Bu$$

These should look familiar from:

- MATH2940 Linear Algebra
- ECE3250 Signals and Systems
- ECE5210 Theory of Linear Systems
- MAE3260 System Dynamics
- and many others...



Review of the Review

- Linear system: $\dot{x} = Ax$
- Solution: $x(t) = e^{At}x(0)$
- Eigenvectors: $T = \begin{bmatrix} \xi_1 & \xi_2 & \dots & \xi_n \end{bmatrix}$

Eigenvalues: $D = \begin{bmatrix} \lambda_1 \\ \lambda_2 \end{bmatrix}$

>>[T,D] = eig(A)

- Linear Transform: AT = TD
- Solution: $e^{At} = e^{TDT^{-1}t}$
- Mapping from x to z to x: $x(t) = Te^{Dt}T^{-1}x(0)$
- **Optimal Control (LQR)** |>>LQR (A, B, Q, R) • Stability in continuous time: $\lambda = a + ib$, stable iff a < 0

 λ_n

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- Discrete time: $x(k + 1) = \tilde{A}x(k)$, where $\tilde{A} = e^{A\Delta t}$
- Stability in discrete time: $\tilde{\lambda}^n = R^n e^{in\theta}$, stable iff R
- Nonlinear systems: $\dot{x} = f(x)$

• Linearization: $\frac{DJ}{Dx}$

- Controllability: $\dot{x} = (A BK)x$ |>>rank(ctrb(A, B))
- Reachability
- **Controllability Gramian**
- Pole Placement |>>place(A,B,poles)

2	<	1

Controllability

- Is the system controllable? \bullet
 - A system is controllable if you call your state x anywhere you want in
 - Matlab >>rank(ctrb(A, B))
- How do we design the control law,

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$$\dot{x} = Ax + Bu \qquad x \in \mathbb{R}$$
$$\dot{x} = Ax - BKx \qquad A \in \mathbb{R}$$
$$\dot{x} = (A - BK)x \qquad u \in \mathbb{R}$$
n steer
n \mathbb{R}^n
New dynamics $B \in \mathbb{R}^n$
$$u? \qquad \underbrace{u}_{k=Ax+Bu}_{k=Ax+Bu} \qquad \underbrace{y=x}_{\text{"full state feedb}}$$

A linear controller (K matrix) can be optimal for linear systems!











Linear Quadratic Regulator

- What are the optimal eigenvalues for our system? •
 - Tradeoff performance and control effort



- R = 0.001 cost of input energy
- Solved using the Ricatti Equation (compute expensive O(n³)
- Matlab >>K = lqr(A, B, Q, R)



Inverted pendulum on a cart The controller works!

Caveats:

- In simulation
- Practical issues:
 - Imperfect models
 - Nonlinear parts: deadband, saturation, etc.
 - Partial state feedback



Linear Systems — where are we?

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Observability



Observability

- Controllability lacksquare
 - Can we steer the system anywhere we want given some control input u?
- Observability
 - Can we estimate any state x, from a time series of measurements y(t)?



$$\dot{x} = Ax + Bu \qquad x \in \mathbb{R}$$
$$u = -Kx$$
$$\dot{x} = (A - BK)x$$









• Observable iff rank(\mathcal{O}) = n

• >>rank(obsv(A,C))

• Iff a system is observable, we can estimate x from y. We can find the best estimates using the observability gramian

• >> [U, S, V] = svd(\mathcal{O})







KF with **PID**



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Why sensor fusion?

- Partial state feedback
- Bad sensors
- Imperfect model
- Slow feedback





noise

Probabilistic Robotics

- Sources of uncertainty
 - Measurements
 - Actions
 - Models
 - States



- Gaussian distributions
 - $[\mu \pm \sigma]$
 - Symmetric
 - Unimodal
 - Sum to "unity"





- observations
 - Assume that posterior and prior belief are Gaussian variables



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Incorporate uncertainty to get better estimates based on both inputs and

 Assume that posterior and prior belief are Gaussian variables

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State estimate: $\mu(t)$

Process noise: Σ_{μ}

State uncertainty: $\Sigma(t)$







- Assume that posterior and prior belief are Gaussian variables
 - Prediction step
 - x(t) = Ax(t-1) + Bu(t) + n, where
 - $\mu_p(t) = A\mu(t-1) + Bu(t)$
 - $\Sigma_p(t) = A\Sigma(t-1)A^T + \Sigma_u$
 - Update step
 - $K_{KF} = \Sigma_p(t)C^T(C\Sigma_p(t)C^T + \Sigma_z)^{-1}$
 - $\mu(t) = \mu_p(t) + K_{KF}(z(t) C\mu_p(t))$
 - $\Sigma(t) = (I K_{KF}C)\Sigma_p(t)$

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State

6

X S

Drio,





0.8

0.6

0.4

0.2







8

Kalman Filter
Function
$$(\mu(t-1), \Sigma(t-1), u(t), z(t))$$

1. $\mu_p(t) = A\mu(t-1) + Bu(t)$
2. $\Sigma_p(t) = A\Sigma(t-1)A^T + \Sigma_u$
3. $K_{KF} = \Sigma_p(t)C^T(C\Sigma_p(t)C^T + \Sigma_z)$
4. $\mu(t) = \mu_p(t) + K_{KF}(z(t) - C\mu_p(t))$
5. $\Sigma(t) = (I - K_{KF}C)\Sigma_p(t)$
6. Return $\mu(t)$ and $\Sigma(t)$

—

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0.2







covariance matrices:







Kalman Filter
$$(\mu(t-1), \Sigma(t-1), u(t), I)$$

1. $\mu_p(t) = A\mu(t-1) + Bu(t)$
2. $\Sigma_p(t) = A\Sigma(t-1)A^T + \Sigma_u$
3. $K_{KF} = \Sigma_p(t)C^T(C\Sigma_p(t)C^T + \Sigma_z)$
4. $\mu(t) = \mu_p(t) + K_{KF}(z(t) - C\mu_p(t))$
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6. Return $\mu(t)$ and $\Sigma(t)$

Example process and measurement noise $\Sigma_{u} = \begin{bmatrix} \sigma_{1}^{2} & 0 \\ 0 & \sigma_{2}^{2} \end{bmatrix}, \ \Sigma_{z} = \sigma_{3}^{2}$ covariance matrices:

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3

4

5

6









8

Kalman Filter vs Bayes Filter

- Bayes Filter
- prior beliefs to speed up computation

Bayes Filter(bel(
$$x_{t-1}$$
), u_t , z
1. for all $x(t)$ do
2. $\overline{bel}(x(t)) = \Sigma(x(t-1))$
3. $bel(x(t)) = \alpha p(z(t))$
4. end for
5. return bel(x_t)

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Kalman Filter uses the same idea, but uses Gaussian variables for posterior and

z_t)

p(x(t) | u(t), x(t-1)) bel(x(t-1))x(t))bel(x(t))

Lab 5-8: PID control — Sensor Fusion — Stunt

- Labs 5 and 6: get basic PID to work, consider sampling time, start slow
- Lab 7: Sensor Fusion (model + ToF to get quick estimates of distance from the wall)
 - Perform a step response with the robot and build the state space equations
 - Estimate covariance matrices for process and sensor noise
 - Try the Kalman Filter in Jupyter with your own data from Lab 5
 - Implement the Kalman Filter on your robot
- Lab 8: Use KF + PID to execute fast stunts



$$F = ma = m\ddot{x}$$
$$F = u - \dot{x}$$
$$m\ddot{x} = u - d\dot{x}$$
$$\dddot{x} = \frac{u}{m} - \frac{d}{m}\dot{x}$$

What are d and m?







$$F = ma = m\ddot{x}$$
$$F = u - \dot{x}$$
$$m\ddot{x} = u - d\dot{x}$$
$$\dddot{x} = \frac{u}{m} - \frac{d}{m}\dot{x}$$

What are d and m?

At constant speed, we can find d:

$$0 = \frac{u}{m} - \frac{d}{m}\dot{x} \qquad d = \frac{u}{\dot{x}}$$





State space equations

$$\begin{bmatrix} \dot{x} \\ \dot{x} \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 0 & -\frac{d}{m} \end{bmatrix} \begin{bmatrix} x \\ \dot{x} \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{m} \end{bmatrix}$$

$$C = \begin{bmatrix} -1 & 0 \end{bmatrix}$$





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What are d and m?

At constant speed, we can find d:

$$0 = \frac{u}{m} - \frac{d}{m}\dot{x} \qquad d = \frac{u}{\dot{x}}$$

(assume u=1 for now)
$$d \approx \frac{1}{2000 \text{ mm/s}}$$





State space equations

$$\begin{bmatrix} \dot{x} \\ \dot{x} \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 0 & -\frac{d}{m} \end{bmatrix} \begin{bmatrix} x \\ \dot{x} \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{m} \end{bmatrix}$$

$$C = \begin{bmatrix} -1 & 0 \end{bmatrix}$$





$$F = ma = m\ddot{x}$$
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$$\dddot{x} = \frac{u}{m} - \frac{d}{m}\dot{x}$$

1st order syste

$$dy(t)$$
 1
 $dt(t)$
 $+$
 dt
 τ

 Unit step respondent

 $y(t) = 1 - e^{-\frac{t}{\tau}}$

What are d and m?

Use the rise time to determine m

$$\dot{v} = \frac{u}{m} - \frac{d}{m}v$$

$$v = 1 - e^{-\frac{d}{m}t_{0.9}} \quad \ln(1 - v) = -\frac{d}{m}t_{0.9}$$

$$m = \frac{-dt_{0.9}}{\ln(1 - 0.9)}$$

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State space equations $\begin{bmatrix} \dot{x} \\ \dot{x} \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 0 & -\frac{d}{m} \end{bmatrix} \begin{bmatrix} x \\ \dot{x} \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{m} \end{bmatrix}$ $C = \begin{bmatrix} -1 & 0 \end{bmatrix}$





$$F = ma = m\ddot{x}$$
$$F = u - \dot{x}$$
$$m\ddot{x} = u - d\dot{x}$$
$$\ddot{x} = \frac{u}{m} - \frac{d}{m}\dot{x}$$

1st order syste

$$\frac{dy(t)}{dt} + \frac{1}{\tau}y(t)$$

$$\frac{1}{\tau}y(t) = 1 - e^{-\frac{t}{\tau}}$$







$$F = ma = m\ddot{x}$$
$$F = u - \dot{x}$$
$$m\ddot{x} = u - d\dot{x}$$
$$\dddot{x} = \frac{u}{m} - \frac{d}{m}\dot{x}$$

1st order syste

$$dy(t)$$
 1

 $dt(t)$
 + $\tau y(t)$
 dt
 τ

 Unit step response

 $y(t) = 1 - e^{-\frac{t}{\tau}}$

What are d and m?

Use the rise time to determine m

$$\dot{v} = \frac{u}{m} - \frac{d}{m}v$$

$$v = 1 - e^{-\frac{d}{m}t_{0.9}} \qquad \ln(1 - v) = -\frac{d}{m}t_{0.9}$$

$$m = \frac{-dt_{0.9}}{\ln(1 - 0.9)} = \frac{-0.0005 \cdot 1.9}{\ln(0.1)} = \frac{-0.0005 \cdot 1.9}{\ln(0.1)}$$

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State space equations $\begin{bmatrix} \dot{x} \\ \dot{x} \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 0 & -\frac{d}{m} \end{bmatrix} \begin{bmatrix} x \\ \dot{x} \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{m} \end{bmatrix}$ $C = \begin{bmatrix} -1 & 0 \end{bmatrix}$





$$F = ma = m\ddot{x}$$
$$F = u - \dot{x}$$
$$m\ddot{x} = u - d\dot{x}$$
$$\dddot{x} = \frac{u}{m} - \frac{d}{m}\dot{x}$$

What are d and m?

At steady state (constant speed) we can find d (assume u=1 for now) $d = \frac{u}{\dot{x}} \approx 0.0005$

We can use the rise time to find m

$$m = \frac{-dt_{0.9}}{\ln(1 - 0.9)} \approx 4.1258 \cdot 10^{-4}$$





State space equations

$$\begin{bmatrix} \dot{x} \\ \dot{x} \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 0 & -\frac{d}{m} \end{bmatrix} \begin{bmatrix} x \\ \dot{x} \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{m} \end{bmatrix}$$

$$C = \begin{bmatrix} -1 & 0 \end{bmatrix}$$





- We have $A, B, C, \Sigma_{\mu}, \Sigma_{\tau}$
- Discretize the A and B matrices
 - x(n + 1) = x(n) + dx
 - $dx/dt = Ax + Bu \iff dx = dt(Ax + Bu)$
 - x(n + 1) = x(n) + dt(Ax(n) + Bu)
 - $x(n+1) = (I + dt \cdot A)x(n) + dt \cdot Bu$ A_d
 - *dt* is our sampling time (0.130s)
- Rescale from unity input to actual input

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State space equations $\begin{vmatrix} 0 & 1 \\ 0 & -\frac{d}{2} \end{vmatrix} \begin{bmatrix} x \\ \dot{x} \end{bmatrix} + \begin{vmatrix} 0 \\ 1 \\ \frac{1}{2} \end{vmatrix}$ $m \$ $C = [-1 \ 0]$





Lab 7: Kalman Filter Implement the Kalman Filter

Kalman Filter (
$$\mu(t - 1), \Sigma(t - 1), u(t), z(t)$$
)
1. $\mu_p(t) = A\mu(t - 1) + Bu(t)$
2. $\Sigma_p(t) = A\Sigma(t - 1)A^T + \Sigma_u$
3. $K_{KF} = \Sigma_p(t)C^T(C\Sigma_p(t)C^T + \Sigma_z)^{-1}$
4. $\mu(t) = \mu_p(t) + K_{KF}(z(t) - C\mu_p(t))$
5. $\Sigma(t) = (I - K_{KF}C)\Sigma_p(t)$
6. Return $\mu(t)$ and $\Sigma(t)$

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Next, determine measurement and process noise

```
f kf(mu,sigma,u,y):
```

```
mu_p = A.dot(mu) + B.dot(u)
sigma_p = A.dot(sigma.dot(A.transpose())) + Sigma_u
sigma_m = C.dot(sigma_p.dot(C.transpose())) + Sigma_z
kkf_gain = sigma_p.dot(C.transpose().dot(np.linalg.inv(sigma_m)))
y_m = y-C.dot(mu_p)
mu = mu_p + kkf_gain.dot(y_m)
sigma=(np.eye(2)-kkf_gain.dot(C)).dot(sigma_p)
```

return mu,sigma



Lab 7: Kalman Filter **Implement the Kalman Filter**

Measurement noise lacksquare

•
$$\Sigma_z = [\sigma_3^2]$$

• $\sigma_3^2 = (20 \text{mm})^2$

Process noise (dependent on sampling rate)

$$\Sigma_{u} = \begin{bmatrix} \sigma_{1}^{2} & 0 \\ 0 & \sigma_{2}^{2} \end{bmatrix}$$
 Sample

- Trust in modeled position:
 - Pos_{stddev} after 1s: $\sqrt{10^2 \cdot \frac{1}{0.13}} = 27.7 \text{mm}$
- Trust in modeled speed:
 - Speed_{stddev} after 1s: $\sqrt{10^2 \cdot \frac{1}{0.13}} = 27.7 \text{mm/s}$







Lab 7: Kalman Filter Implement the Kalman Filter

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4. $\mu(t) = \mu_p(t) + K_{KF}(z(t) - C\mu_p(t))$
5. $\Sigma(t) = (I - K_{KF}C)\Sigma_p(t)$
6. Return $\mu(t)$ and $\Sigma(t)$

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Finally, determine your initial state mean and covariance

$$\mu(t-1)$$
$$\Sigma(t-1)$$

f kf(mu,sigma,u,y):

```
mu_p = A.dot(mu) + B.dot(u)
    sigma_p = A.dot(sigma.dot(A.transpose())) + Sigma_u
    sigma_m = C.dot(sigma_p.dot(C.transpose())) + Sigma_z
    kkf_gain = sigma_p.dot(C.transpose().dot(np.linalg.inv(sigma_m)))
    y_m = y-C.dot(mu_p)
    mu = mu_p + kkf_gain.dot(y_m)
sigma=(np.eye(2)-kkf_gain.dot(C)).dot(sigma_p)
```

return mu,sigma







