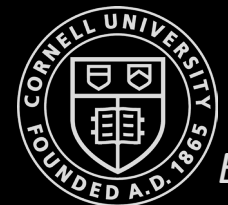
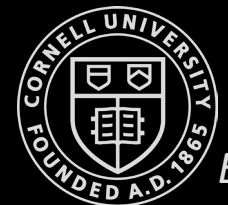
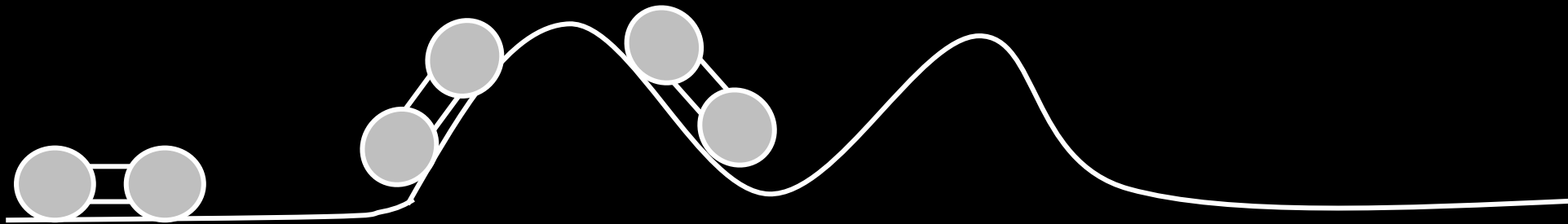


# Fast Robots



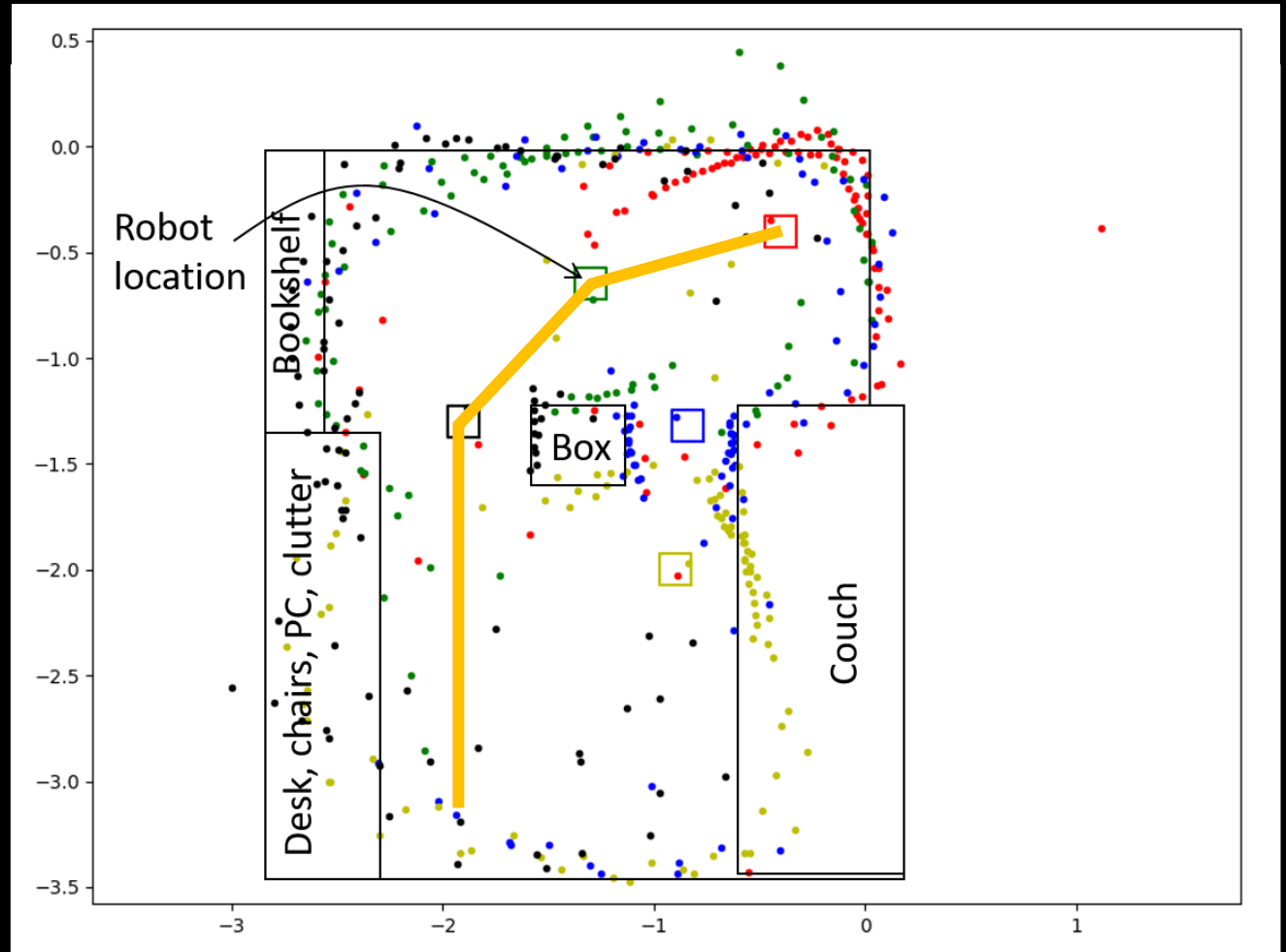
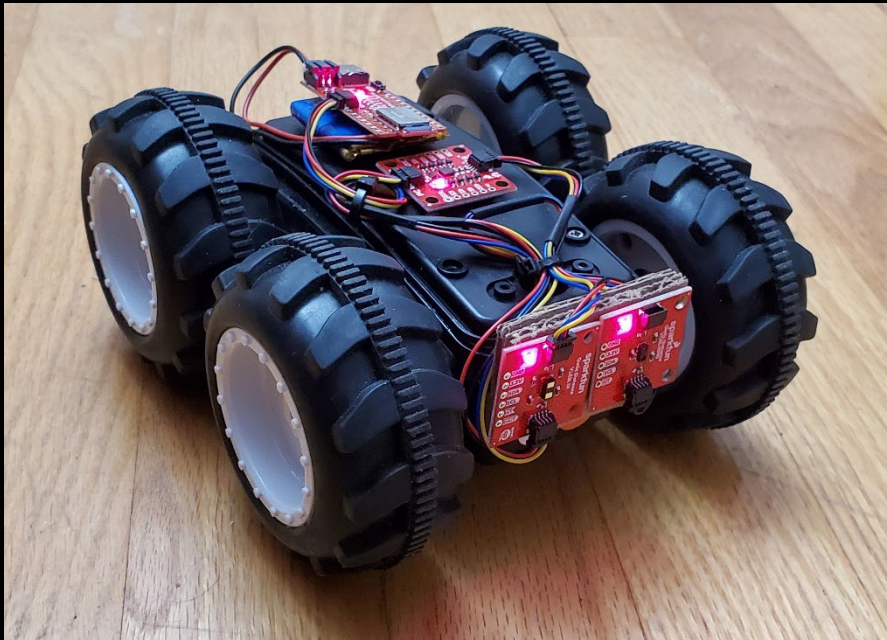
# Feedback Control

- Maintaining speed prediction at different battery levels, over different surfaces
- Maintaining position with respect to walls
- Etc.



# Feedback Control

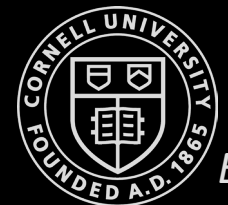
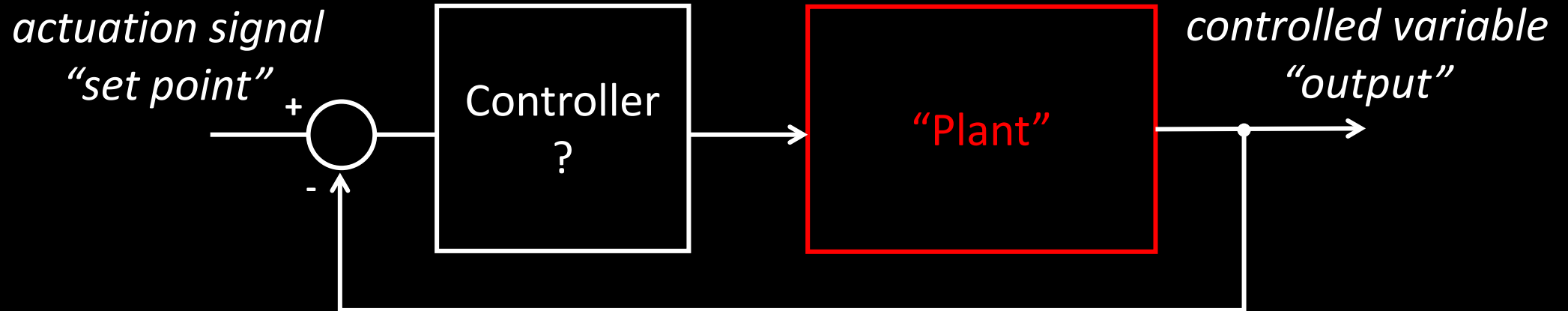
- Maintaining speed prediction at different battery levels and over different surfaces
- Mapping: evenly spaced out sensor readings
- Path execution: adhere to generated path plans



# PID control

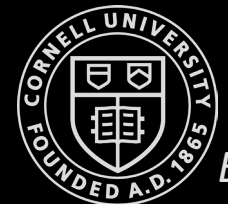
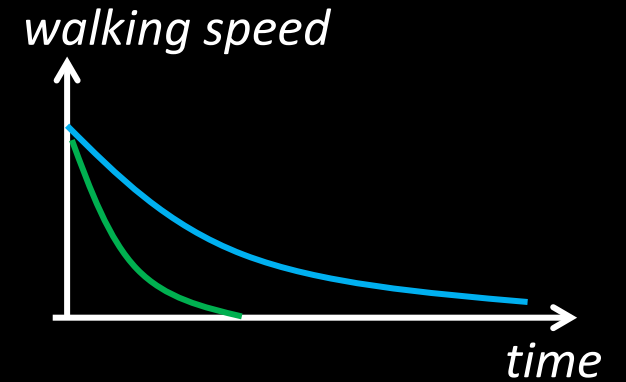
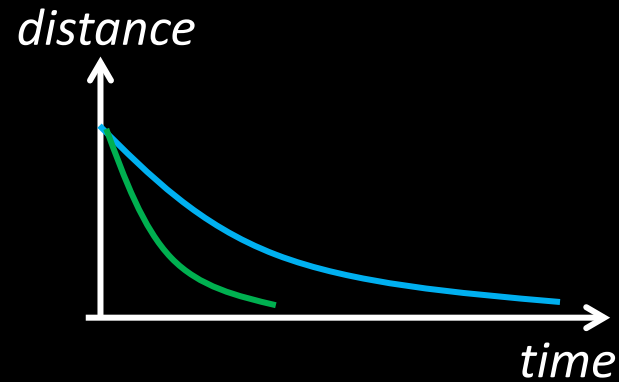
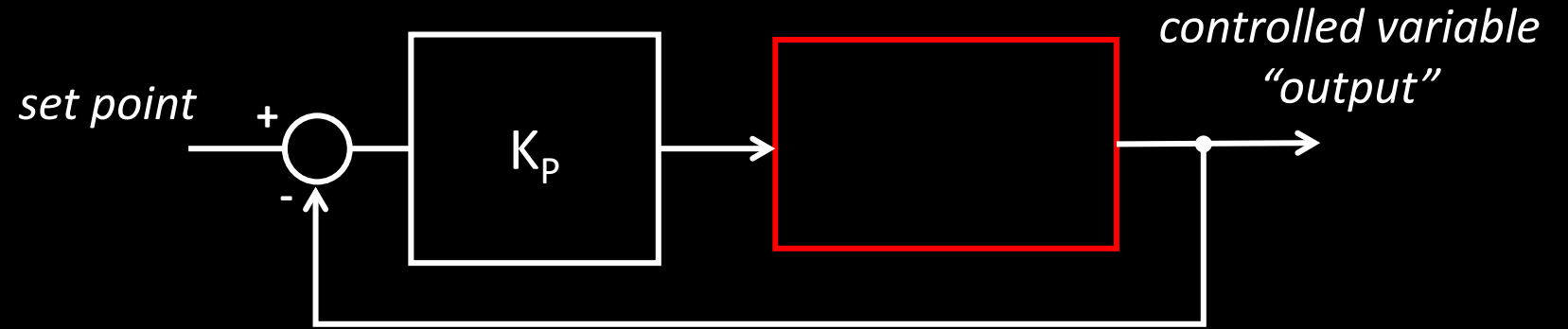
*Heavily inspired by a  
Matlab Tech Talk:  
Understanding PID Control*

$$u(t) = K_P e(t) + K_I \int_0^t e(t) dt + K_D \frac{de(t)}{dt}$$



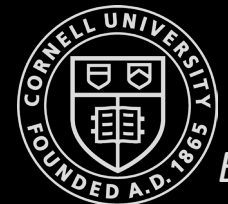
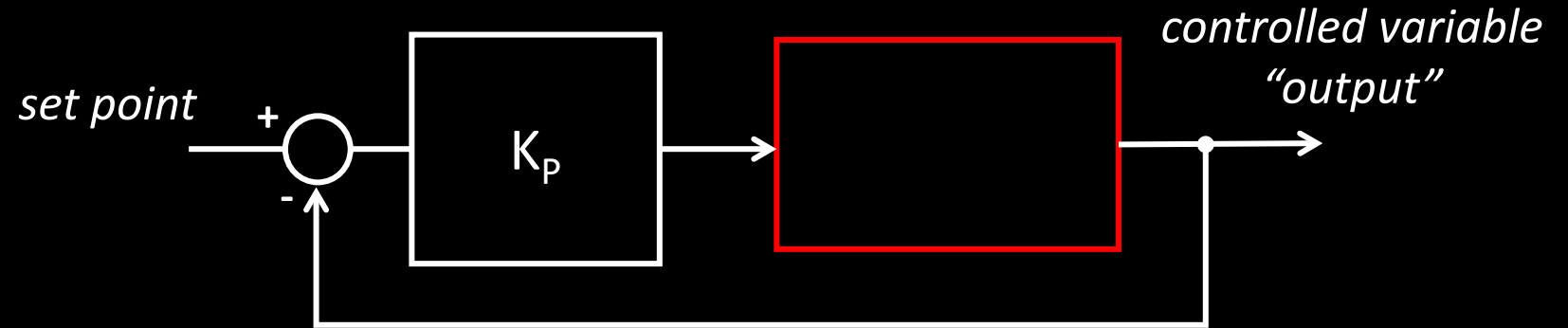
# PID control

- Soccer field example



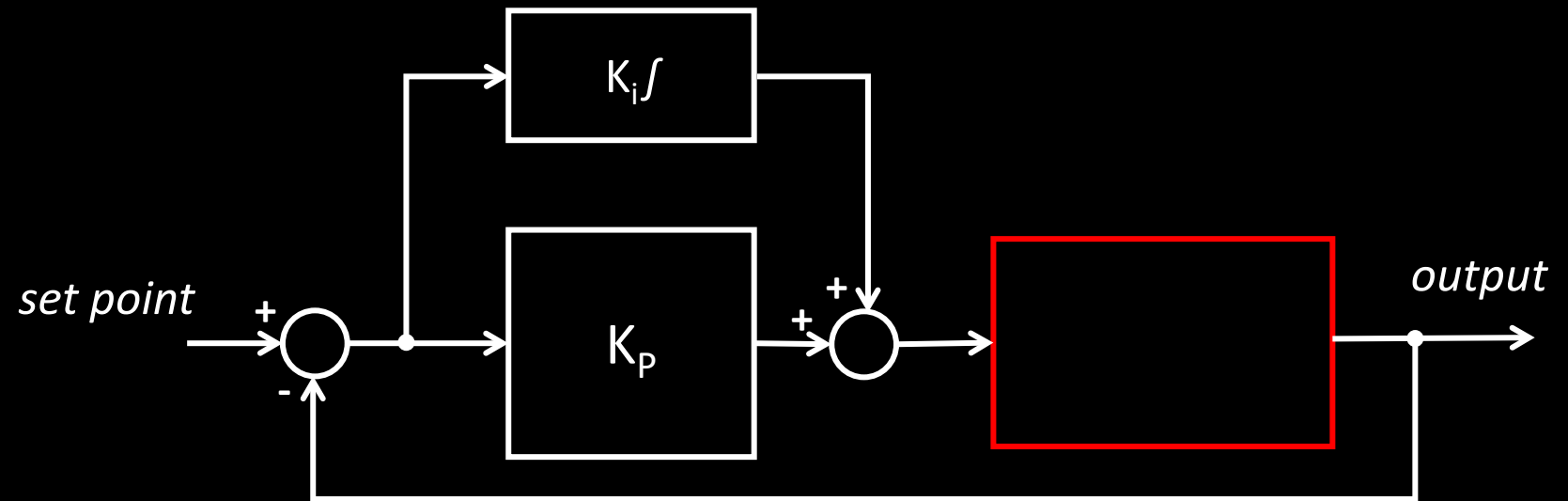
# PID control

- Drone example
  - But there's gravity...
  - Hover at 100rpm
    - $K_p = 2, a > 0m$
    - $K_p = 5, a = 30m$
    - $K_p = 10, a = 40m$
    - $K_p = 100, a = 49m$
    - Steady state error

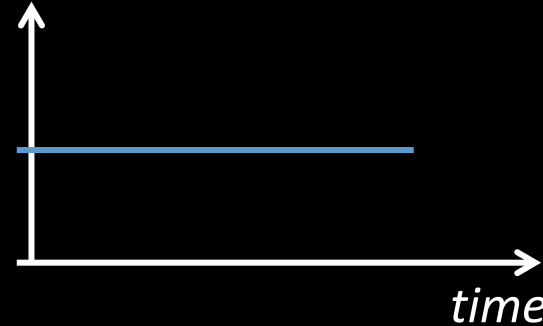


# PID control

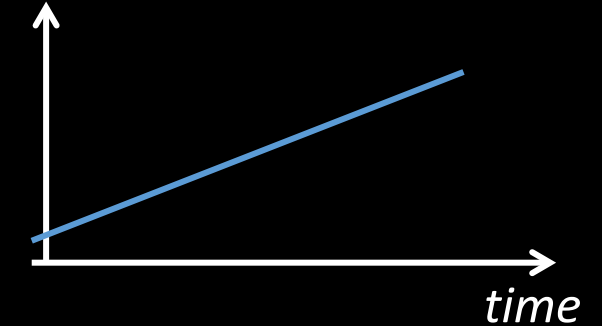
- Drone example
  - But there's gravity...
  - Hover at 100rpm
    - $K_p = 2, a > 0m$



*steady state error*

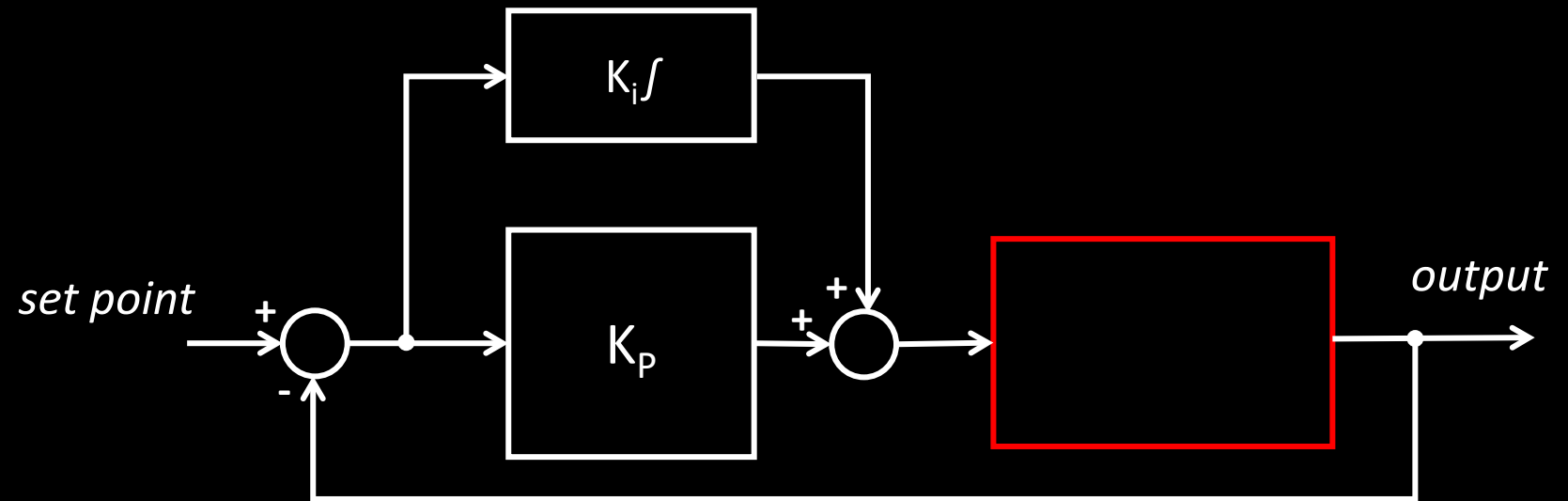


*Integrator output*

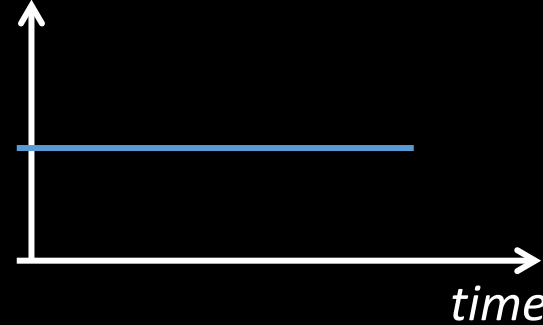


# PID control

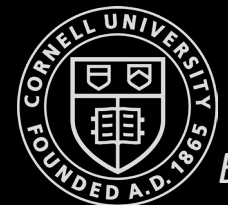
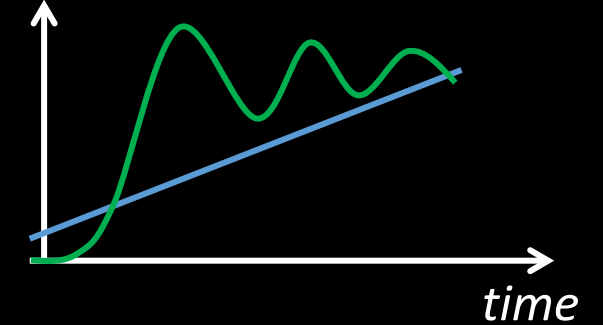
- Drone example



*steady state error*



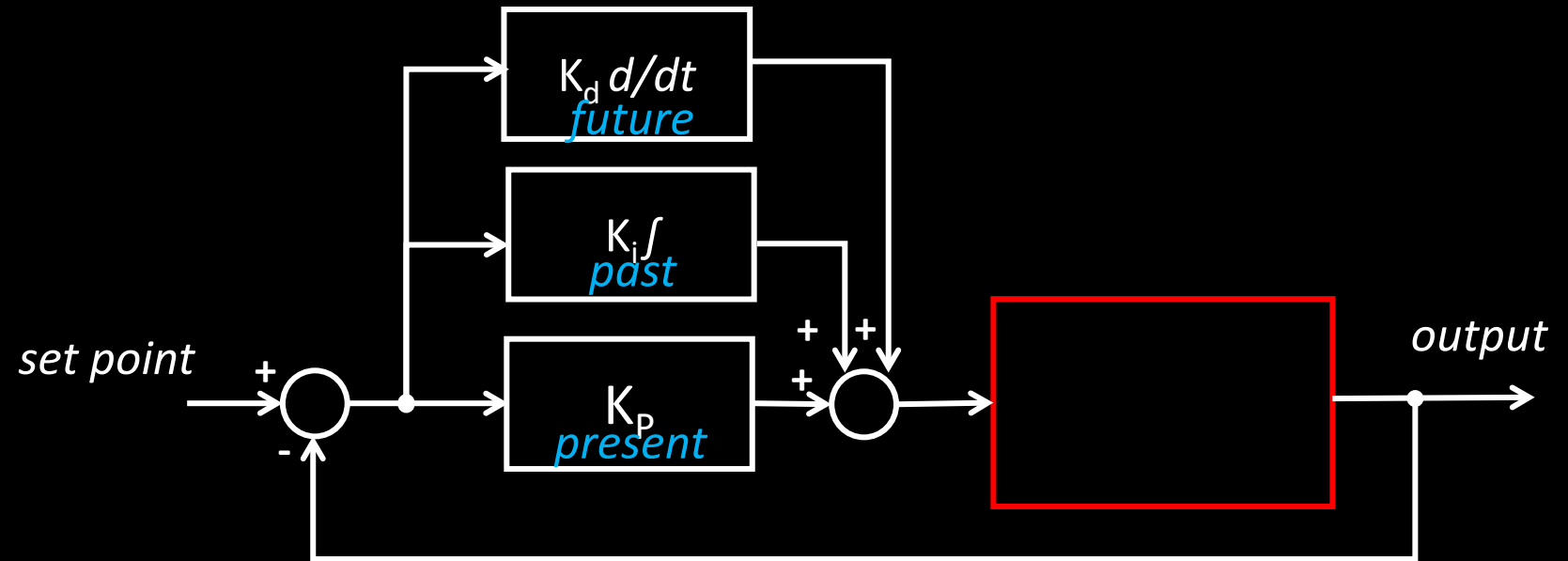
*Integrator output*



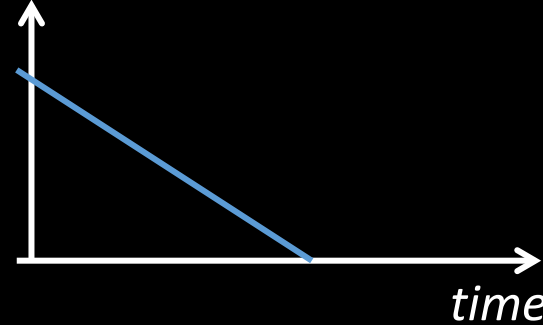


# PID control

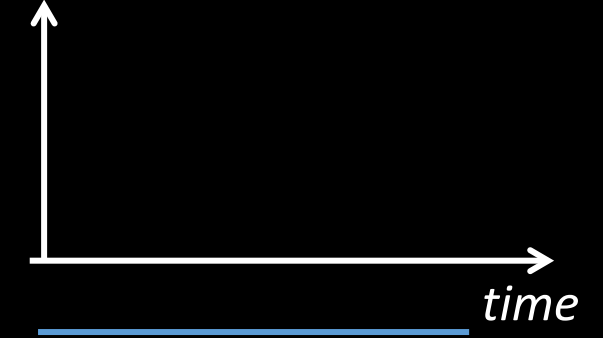
- Drone example



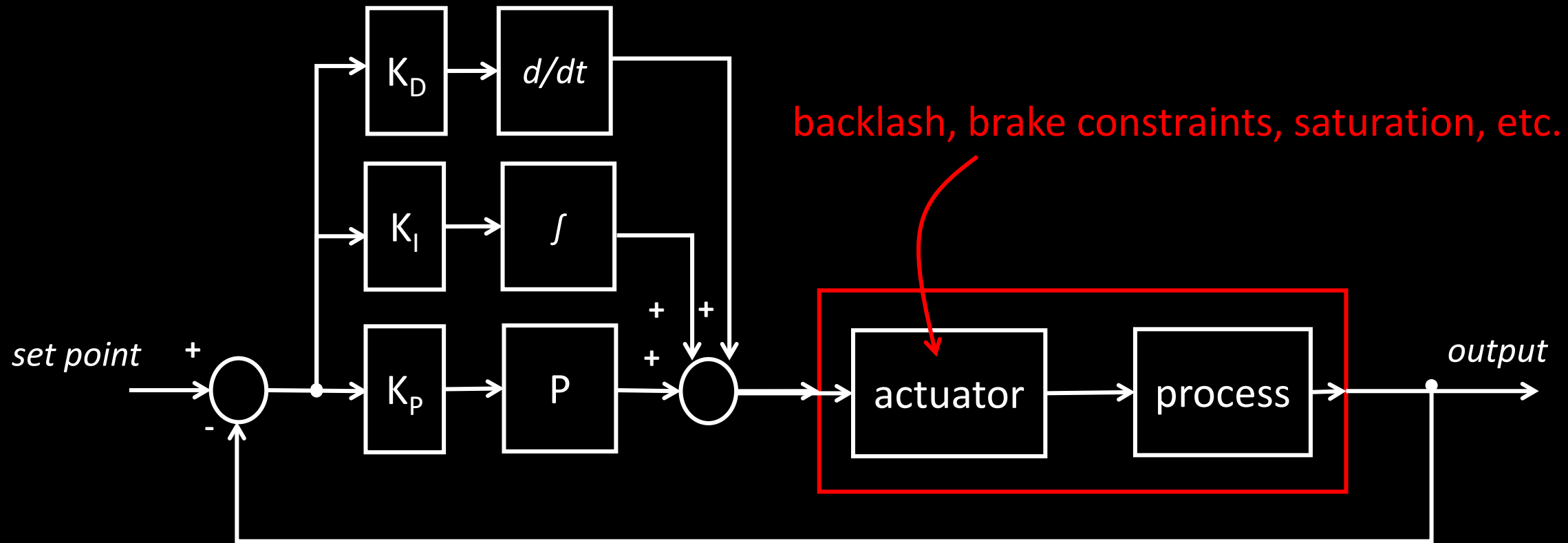
*steady state error*



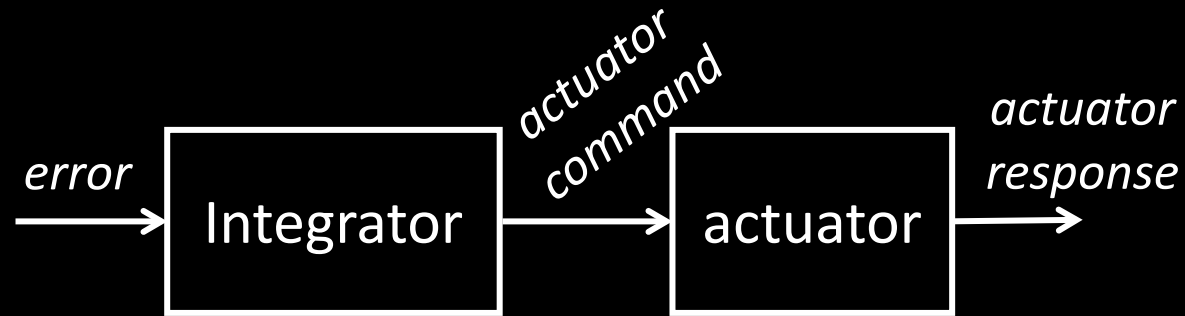
*Derivative output*



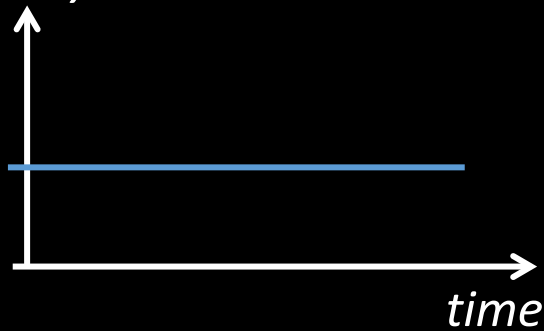
# Real Systems are not linear!



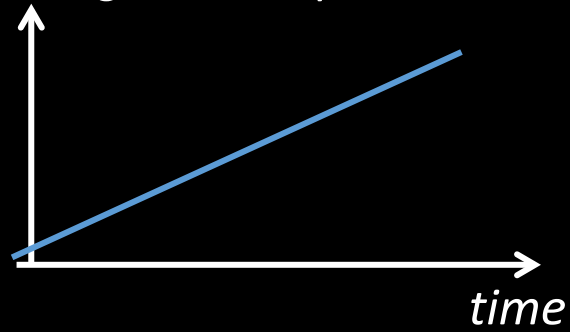
# Real Systems are not linear!



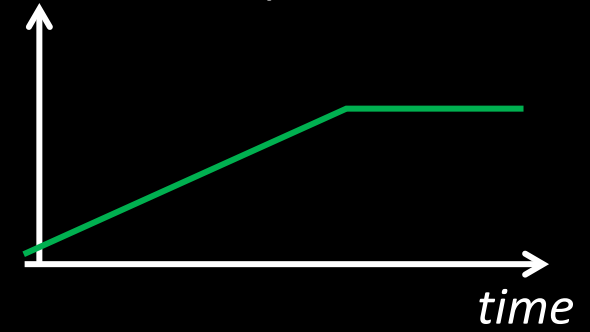
steady state error



Integrator output

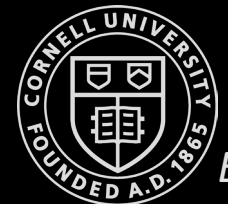
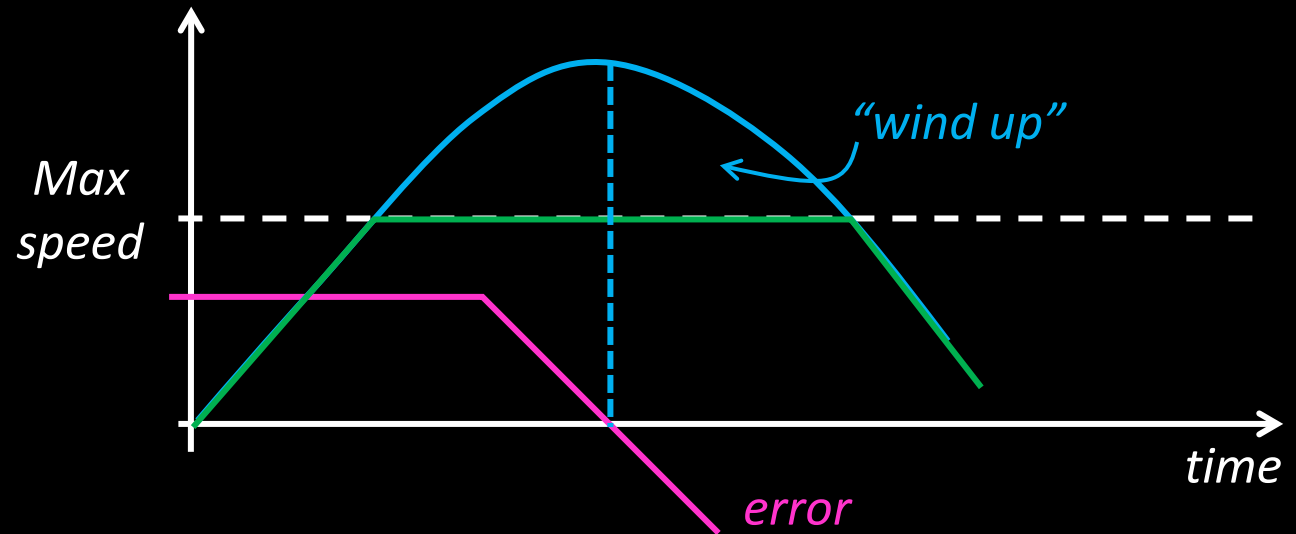
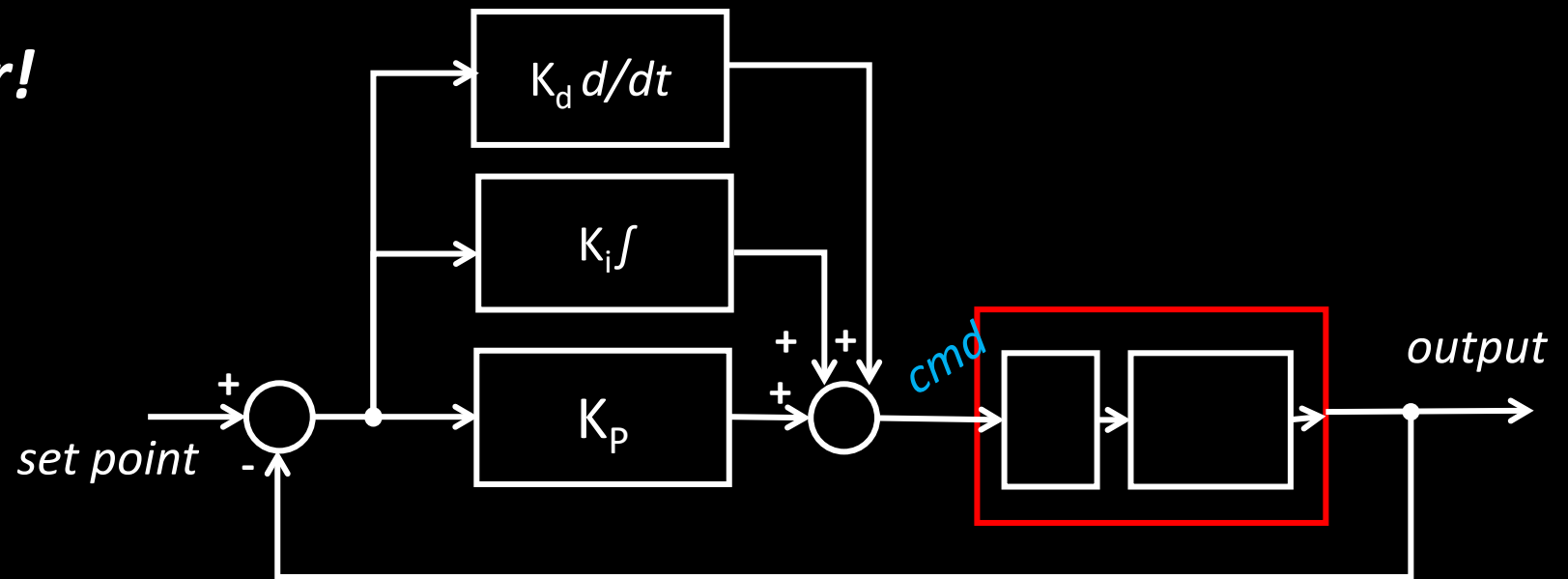


Actuator output



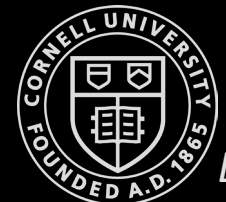
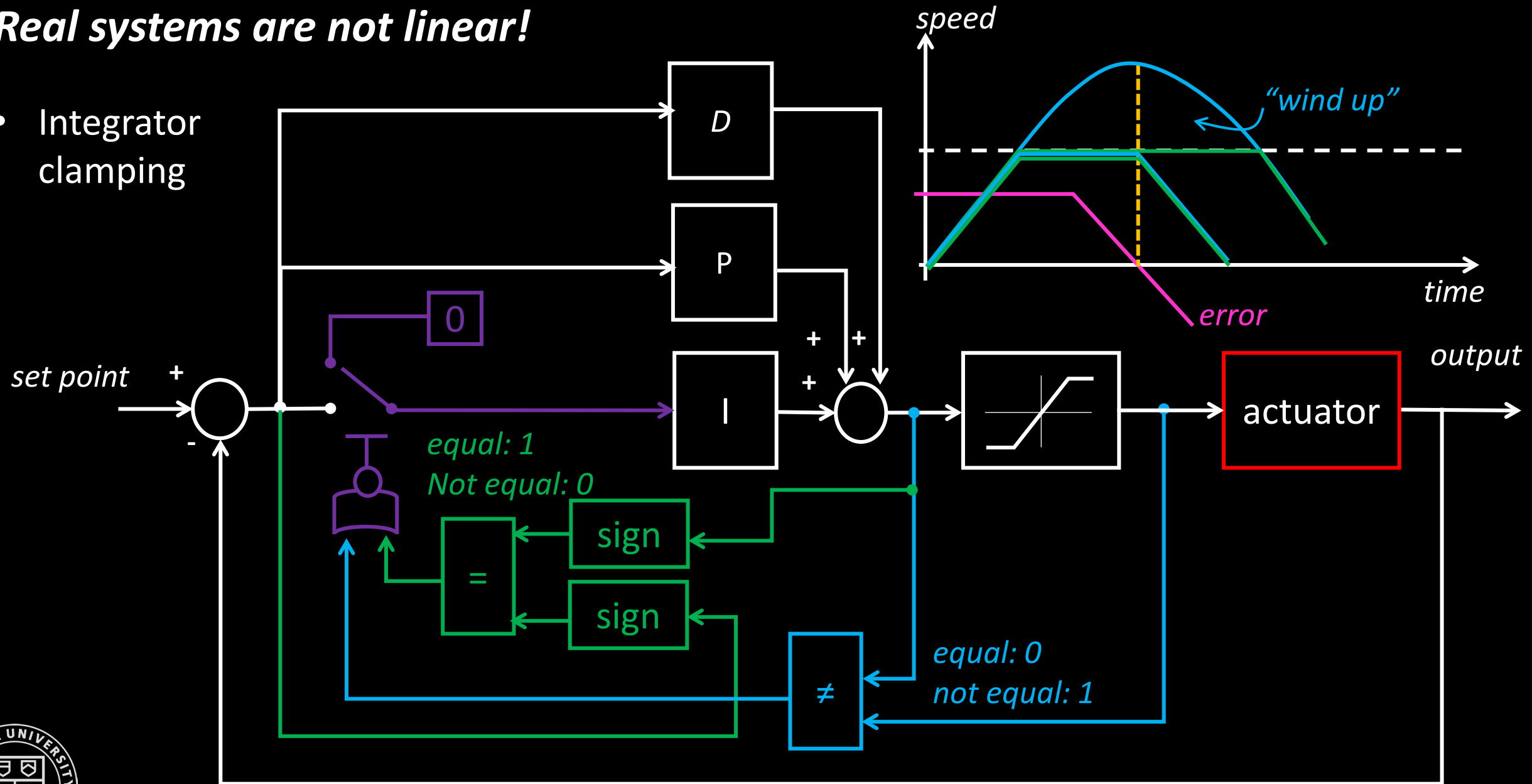
# Real systems are not linear!

- Drone example
  - “Integral wind-up”
  - Clamping



# Real systems are not linear!

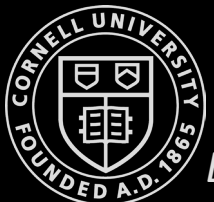
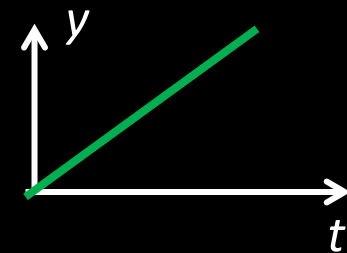
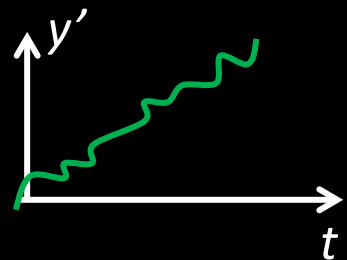
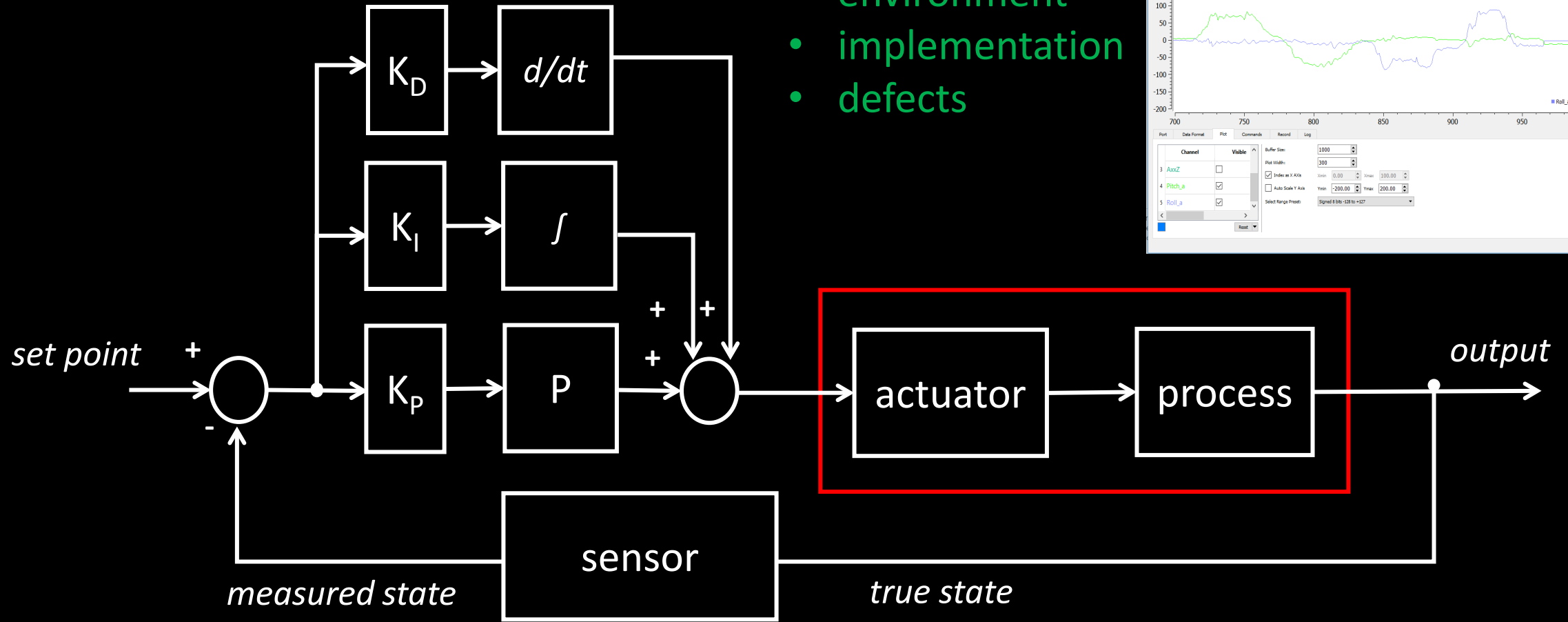
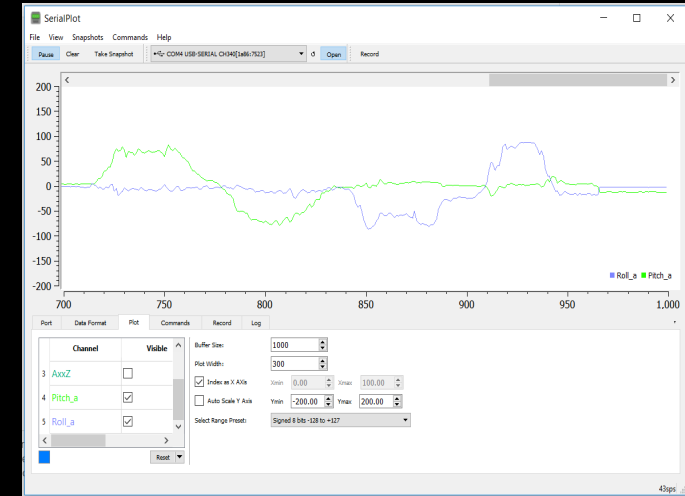
- Integrator clamping



# PID and Sensor noise

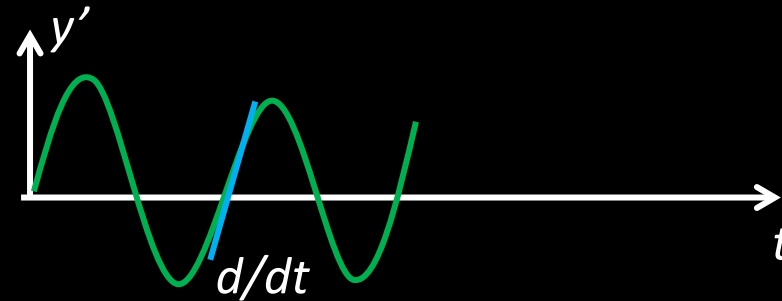
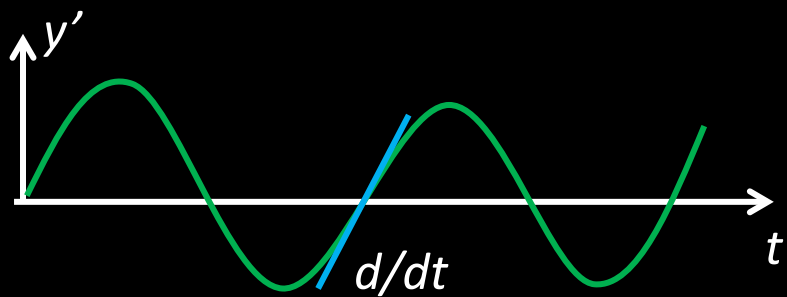
“noise”:

- environment
- implementation
- defects



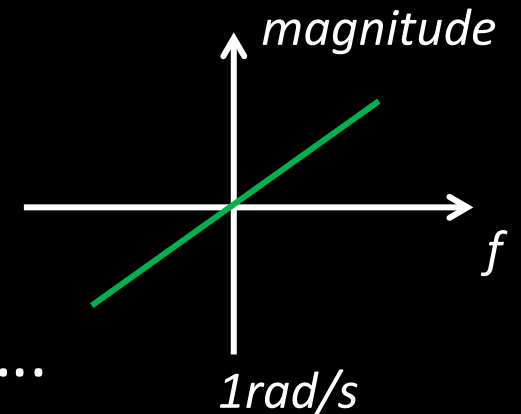
# PID and Sensor noise

- Derivatives amplify HF signals more than LF signals

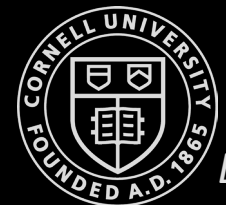


$$y(t) = A\sin(\omega_a t + \phi_a) + B\sin(\omega_b t + \phi_b) + \dots$$

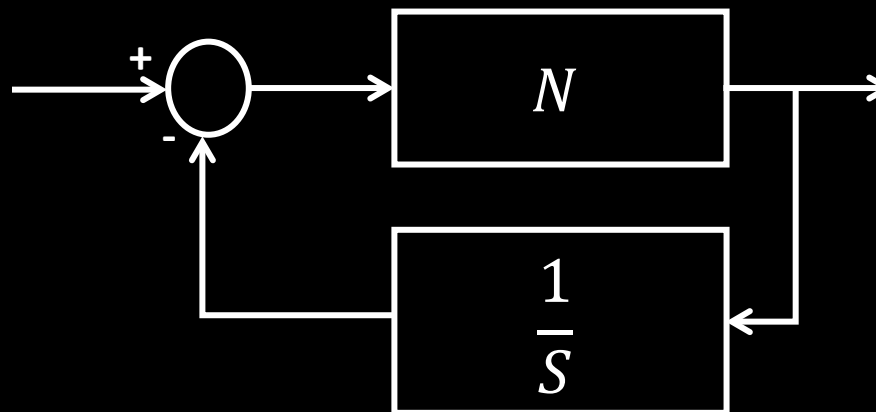
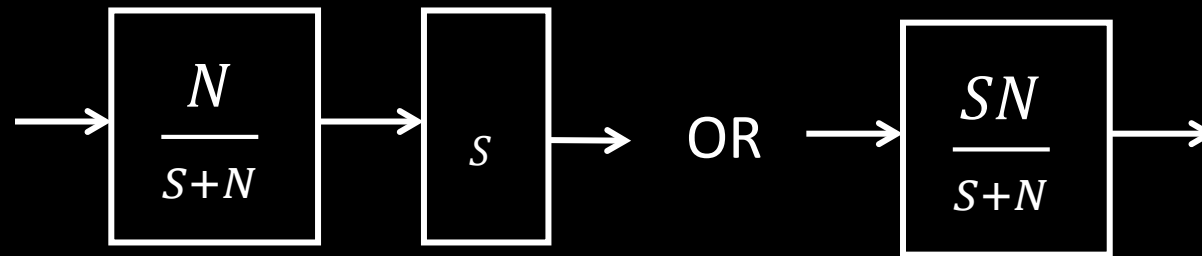
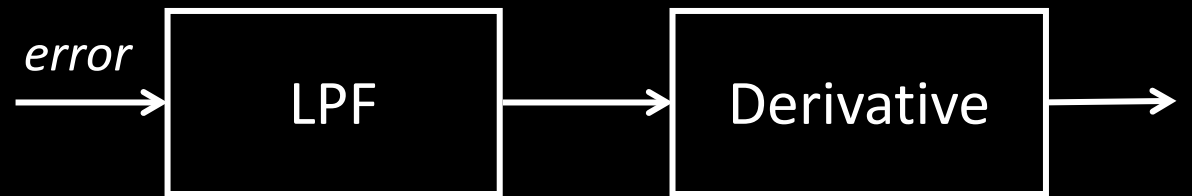
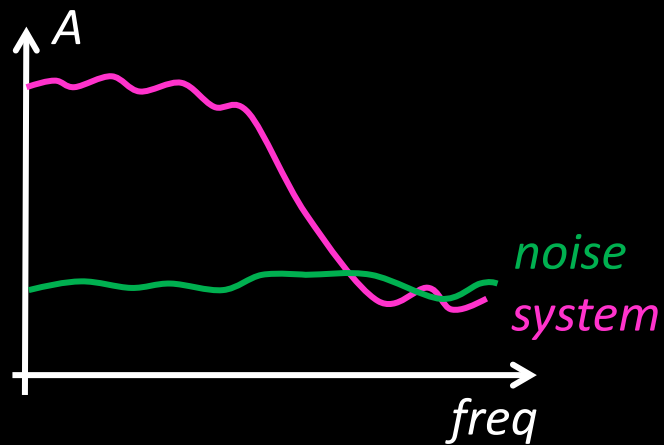
$$dy(t)/dt = A\omega_a \sin(\omega_a t + \phi_a + 90^\circ) + B\omega_b \sin(\omega_b t + \phi_b + 90^\circ) + \dots$$



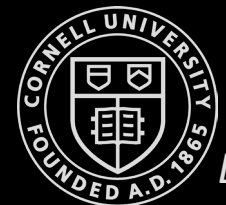
- if  $\omega_a > 1\text{rad/s}$ , the amplitude will increase
- if  $\omega_a < 1\text{rad/s}$ , the amplitude will decrease



# PID and Sensor noise



Time	Laplace
$\frac{d}{dt}$	$S$
$\int dt$	$\frac{1}{S}$
1 <sup>st</sup> order LPF	$\frac{N}{S+N} = \frac{1}{\frac{1}{N}S+1} = \frac{1}{\tau S+1}$





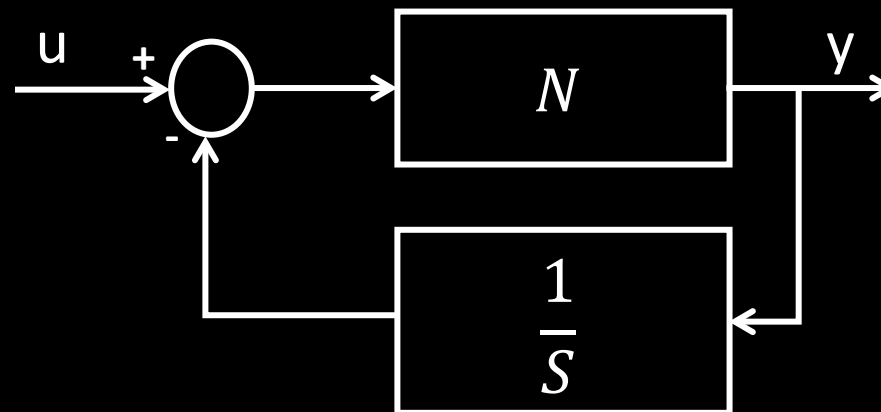
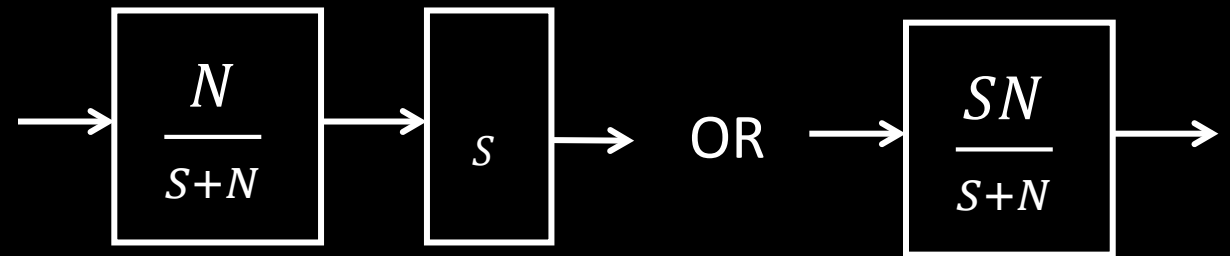
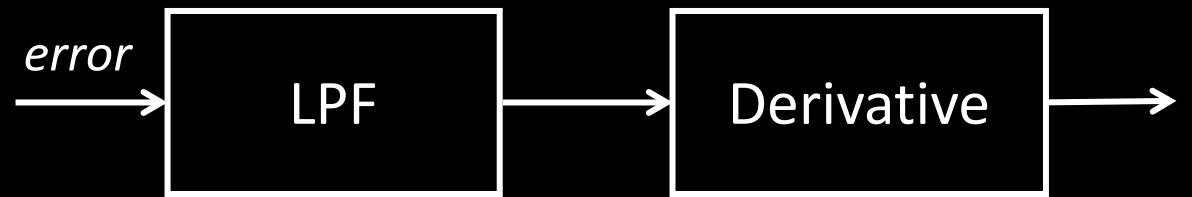
# PID and Sensor noise

$$y = N \left( u - \frac{y}{s} \right)$$

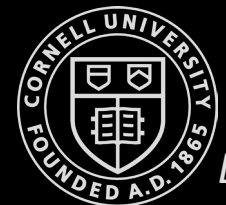
$$y + \frac{Ny}{s} = Nu$$

$$y = \frac{N}{1 + \frac{N}{s}} u$$

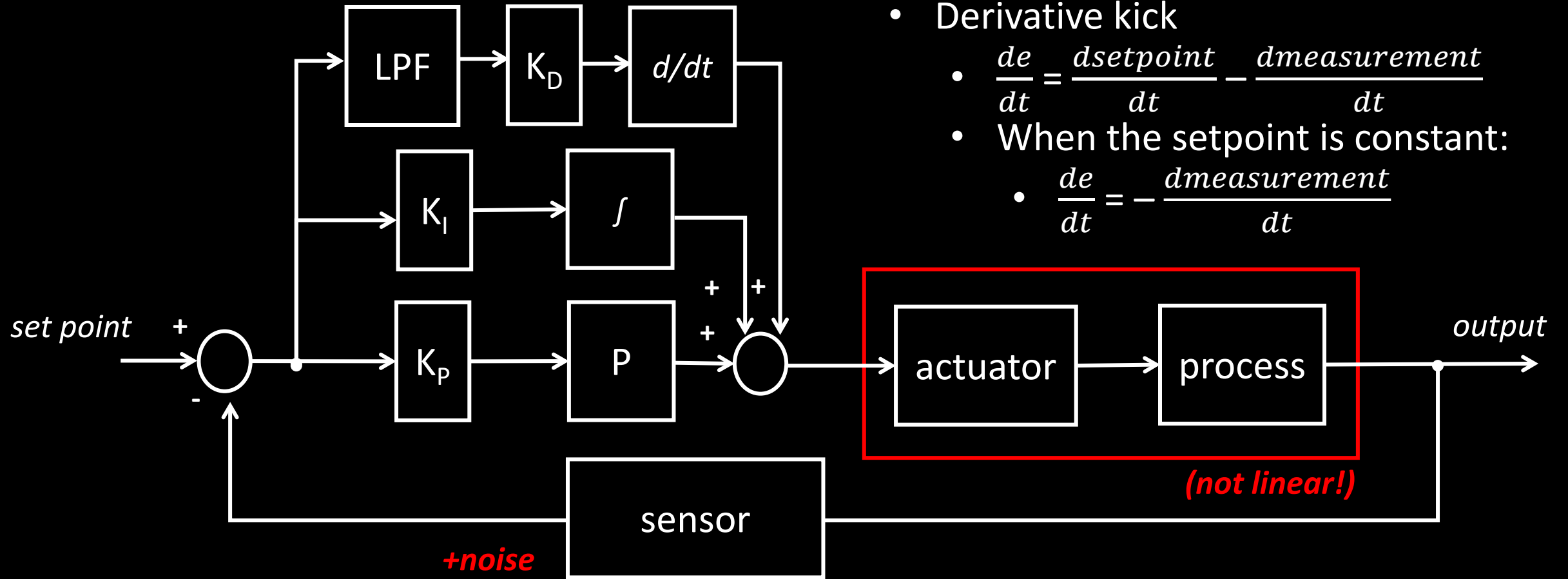
$$\frac{y}{u} = \frac{N}{1 + N\frac{1}{s}}$$



Time	Laplace
$\frac{d}{dt}$	$s$
$\int dt$	$\frac{1}{s}$
1 <sup>st</sup> order LPF	$\frac{N}{S+N} = \frac{1}{\frac{1}{N}S+1} = \frac{1}{\tau S+1}$



# PID



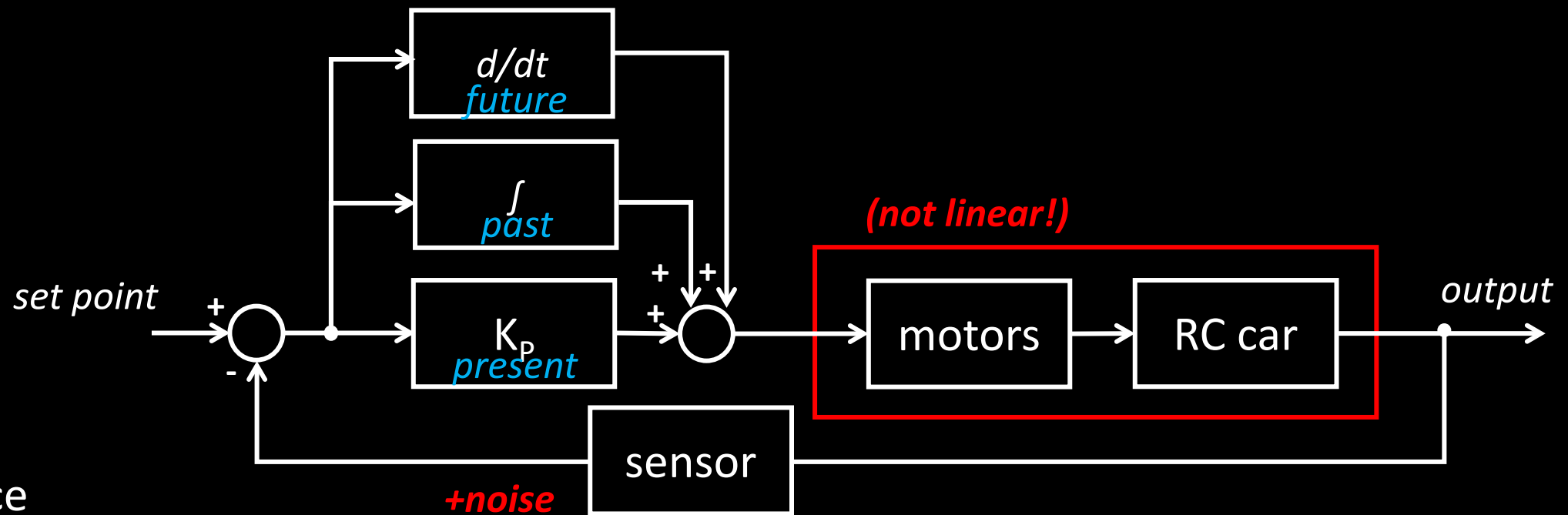
- Integrator wind-up
- Derivative low pass filter
- Derivative kick

- $\frac{de}{dt} = \frac{d\text{setpoint}}{dt} - \frac{d\text{measurement}}{dt}$

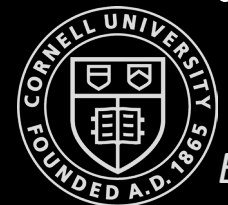
- When the setpoint is constant:

- $\frac{de}{dt} = - \frac{d\text{measurement}}{dt}$

# PID control

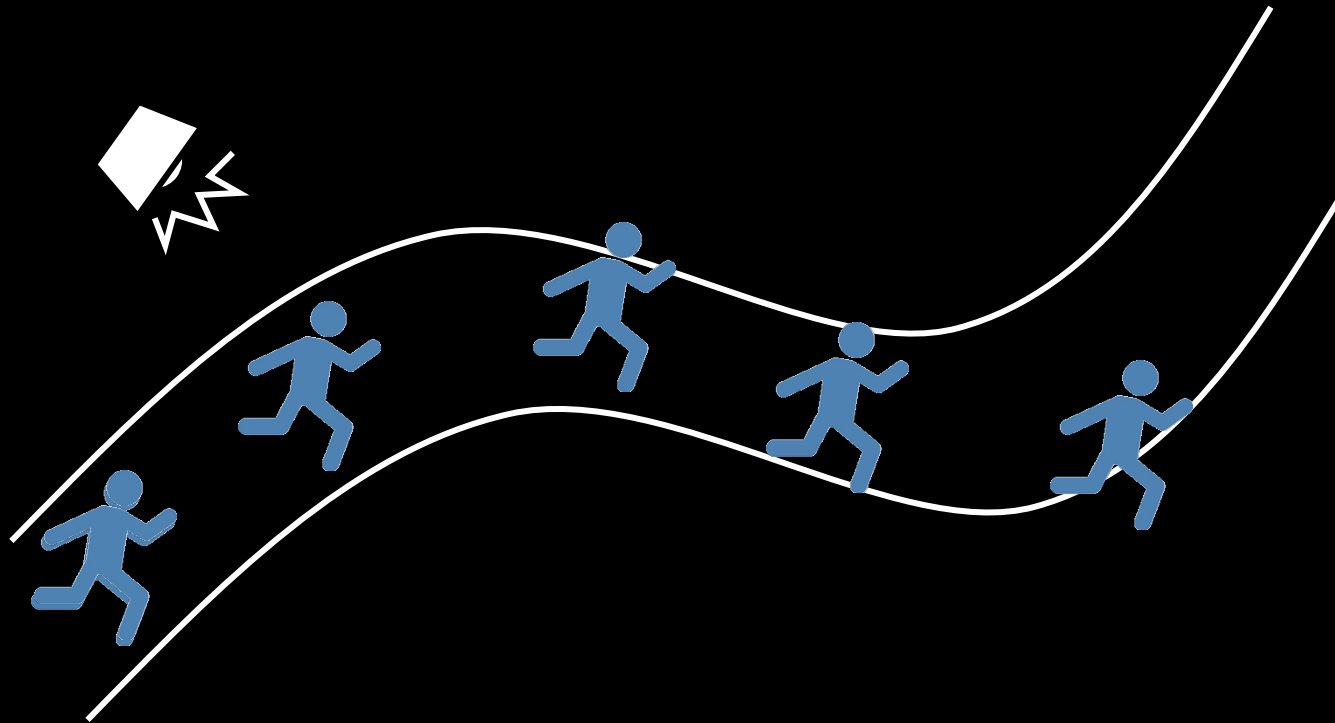


- Performance
  - Rise time/Response
    - Ex: 10% to 90% of final value
  - Peak time
    - Time to reach first peak
  - Overshoot
    - Amount in excess of final value
  - Settling time
    - Ex: Time before output settles to 1% of final value

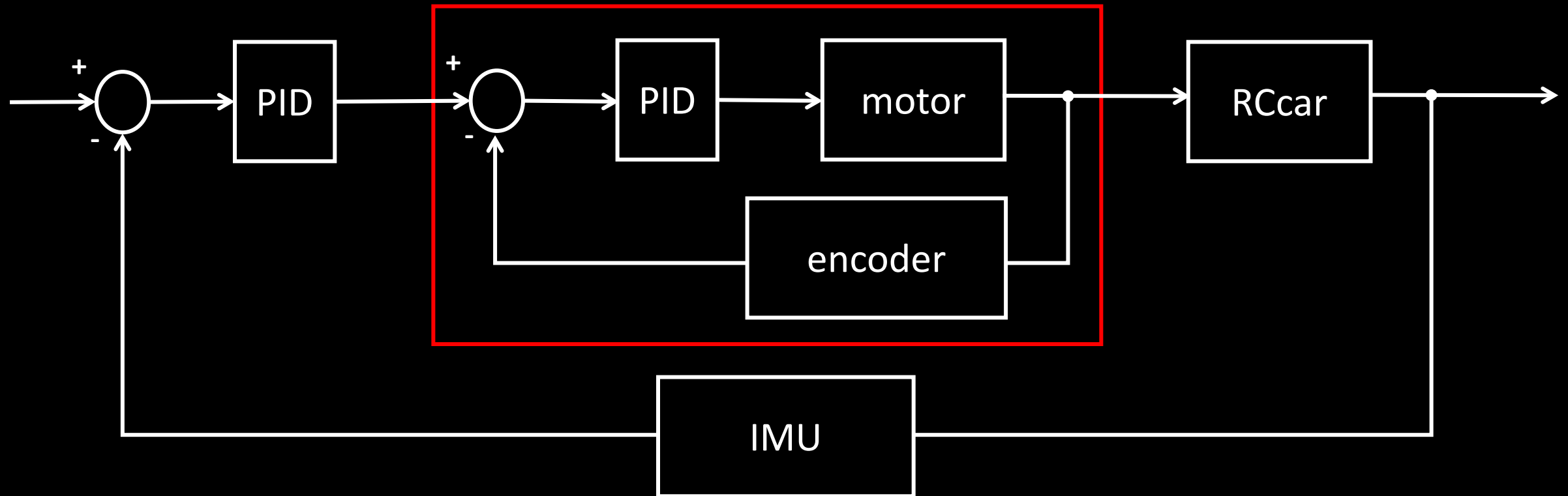


# Discrete PID Control

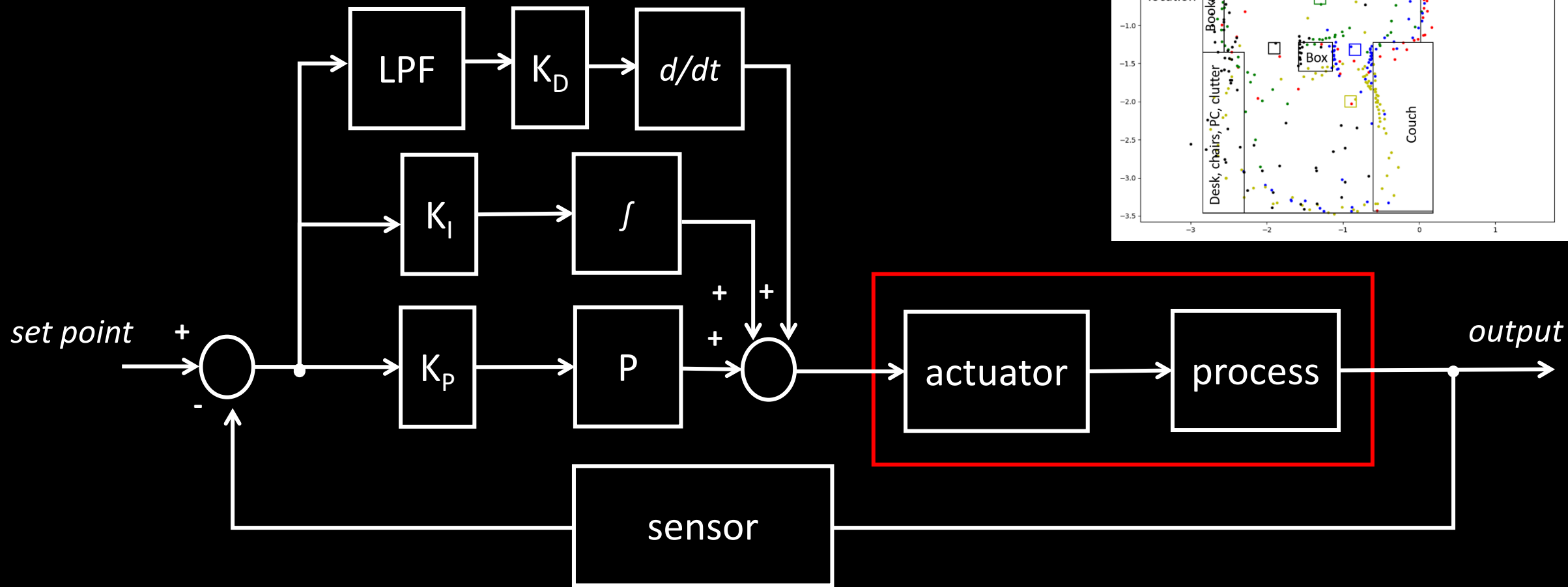
- Sampling time
- Control  $\sim 10$  times faster than the system



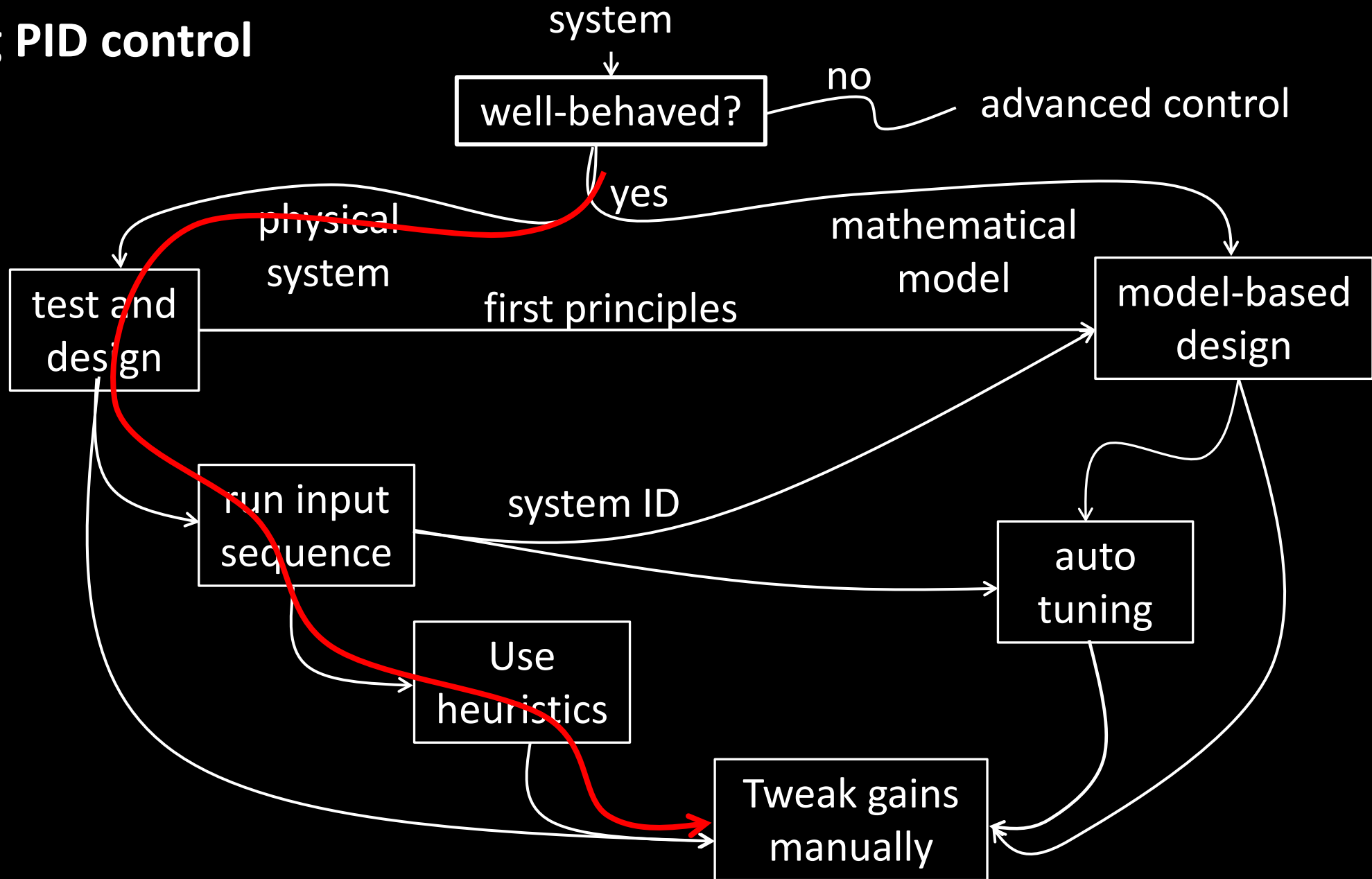
# Cascaded Control Loops



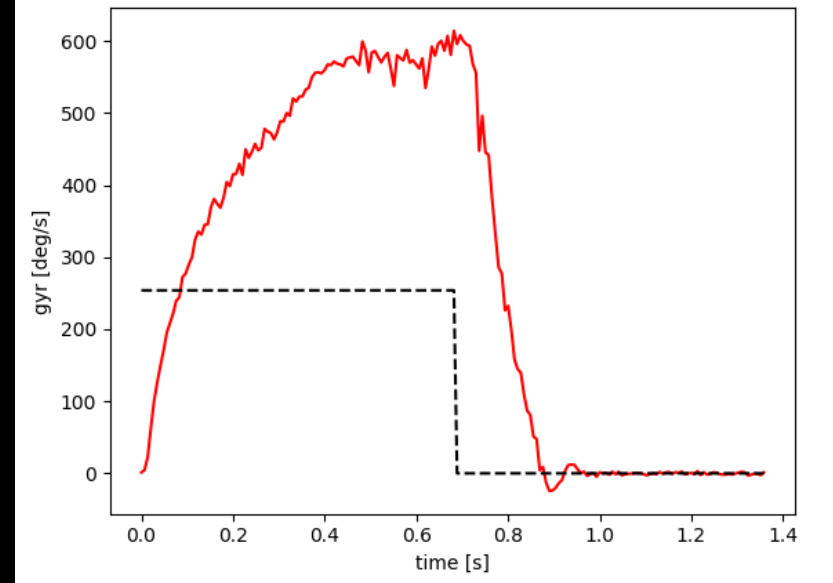
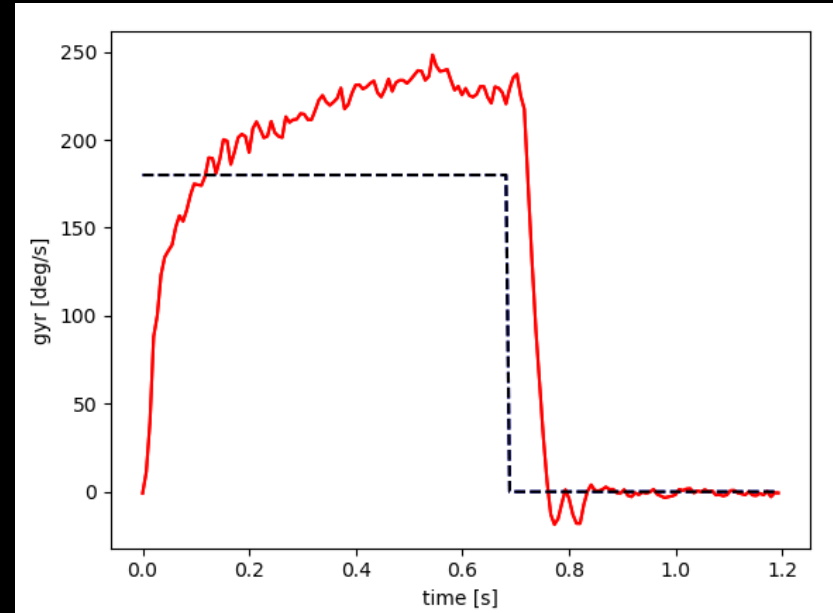
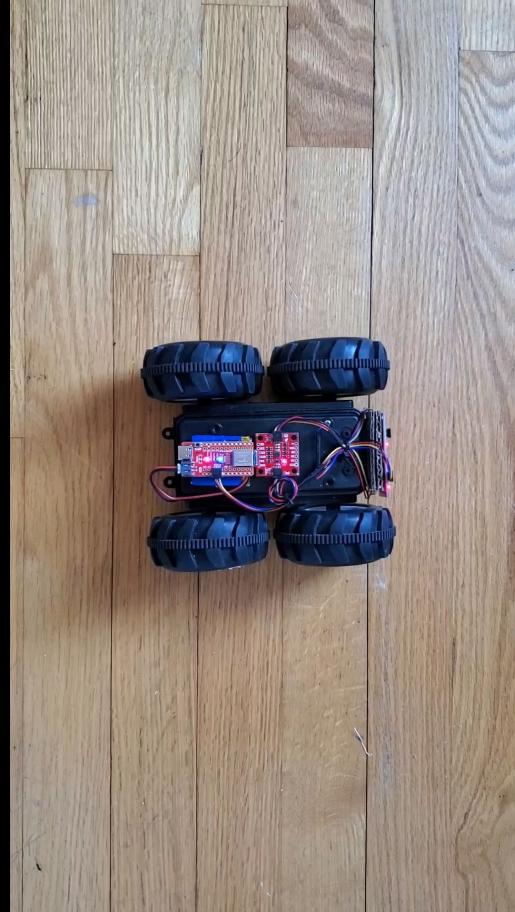
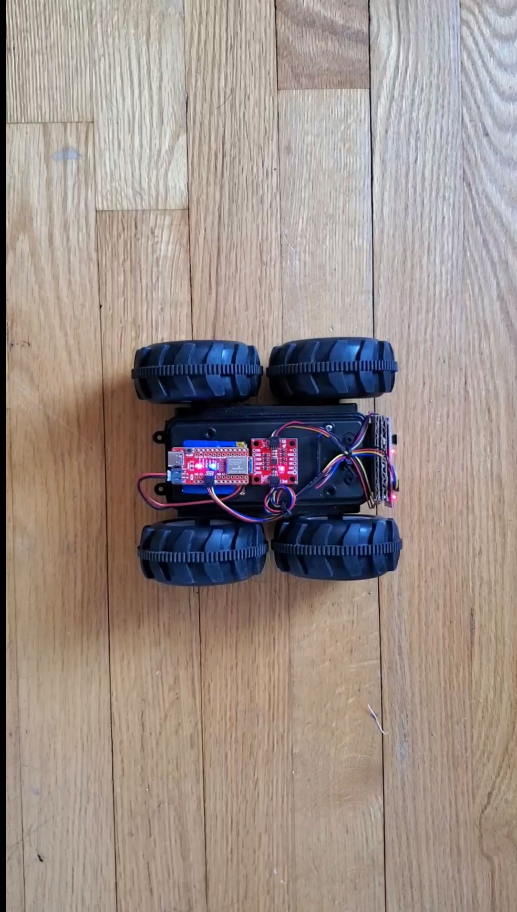
# PID



# Tuning PID control



# Tuning PID control





# Tuning PID control

- Chien, Hornes, and Reswick method

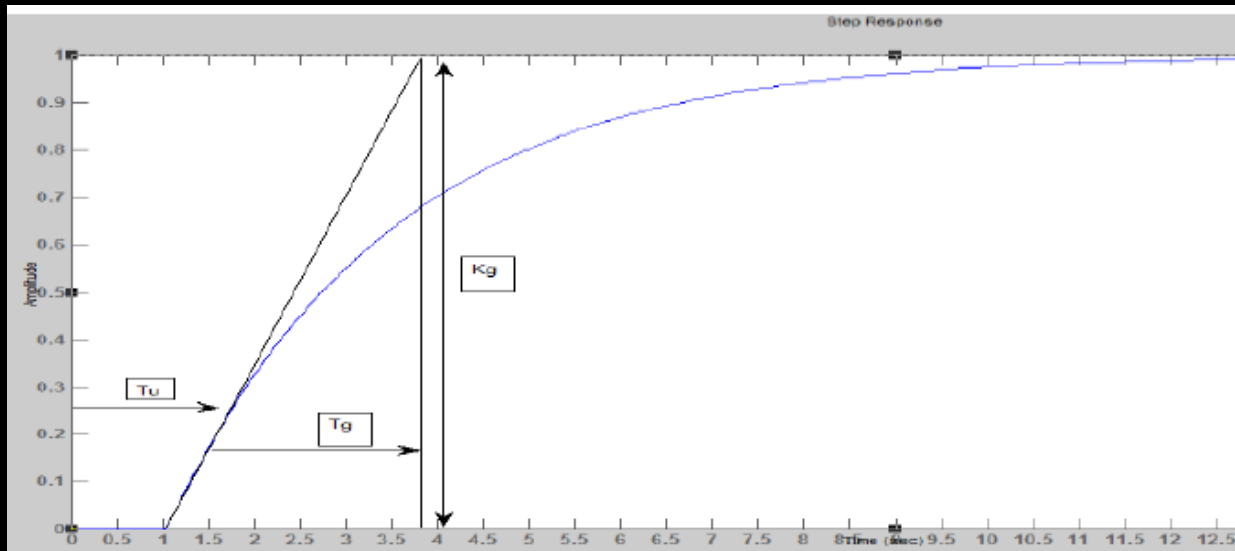
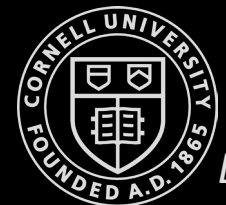
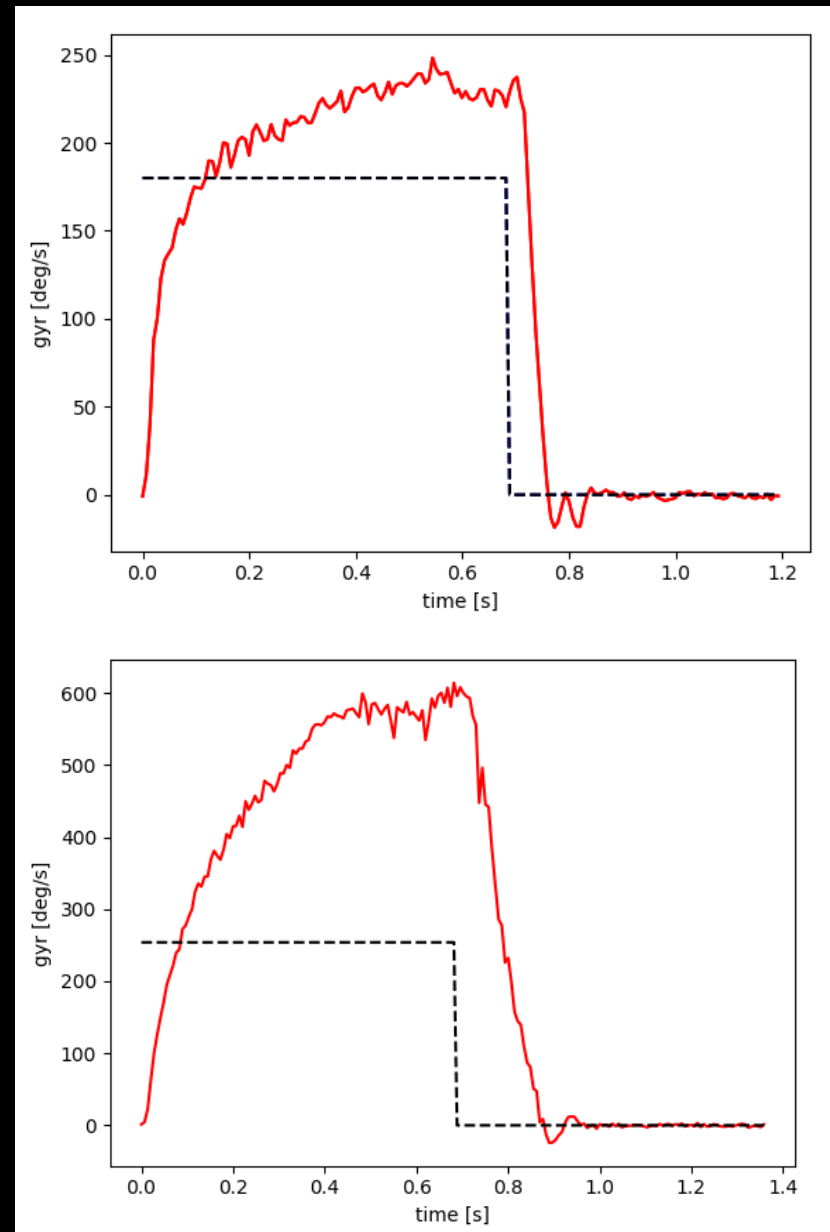


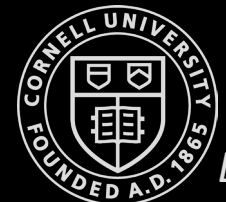
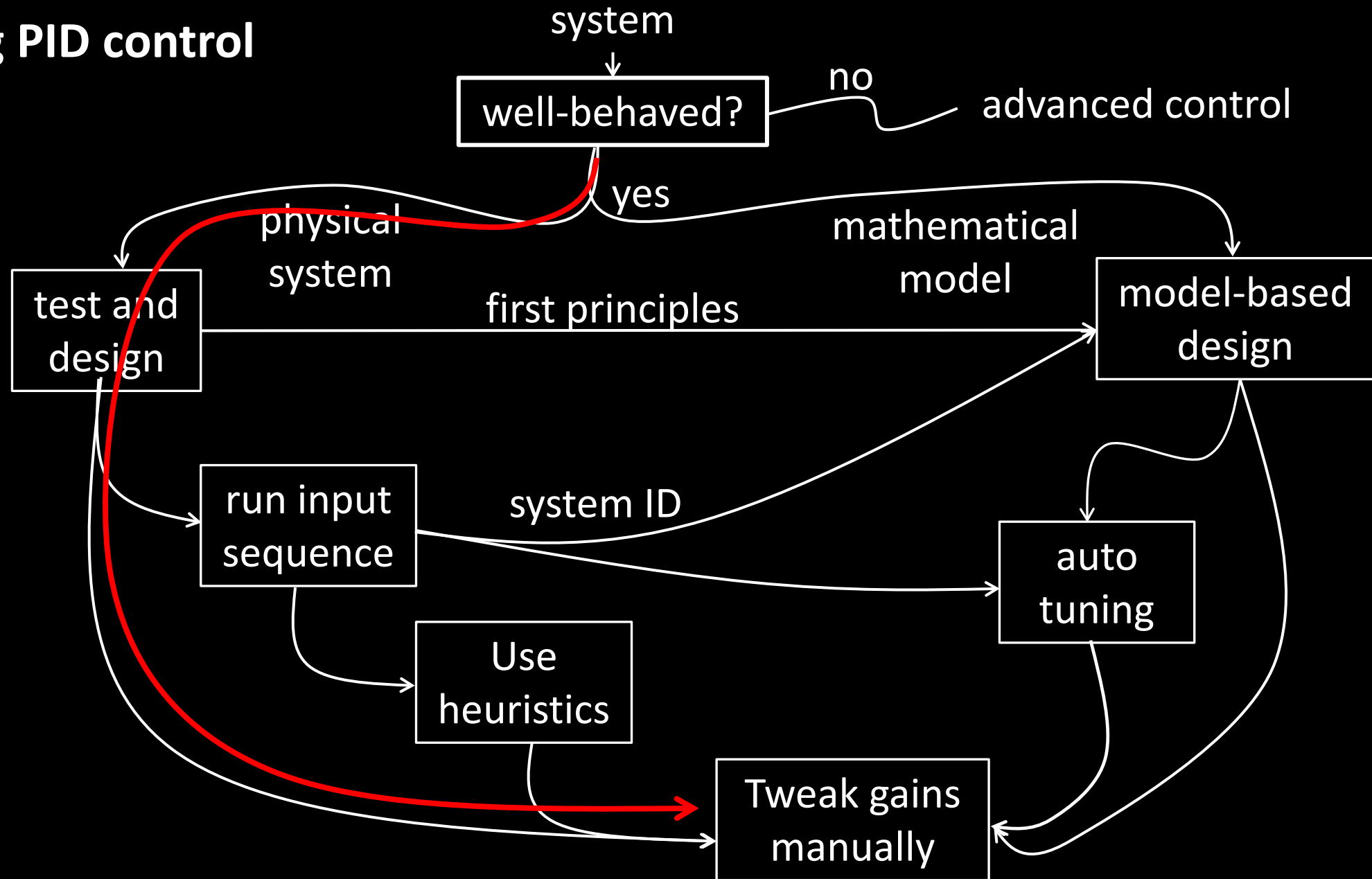
Fig.7. Open loop response of CHR method

Table.11. CHR Compensator

Type of controller	$K_p$	$T_i$	$T_d$
PID	$0.6T_g/T_uK_g$	$T_g$	$0.5T_u$

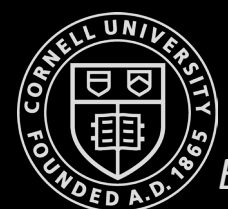


# Tuning PID control

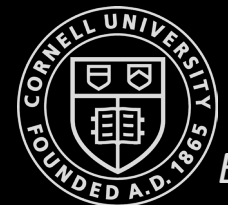
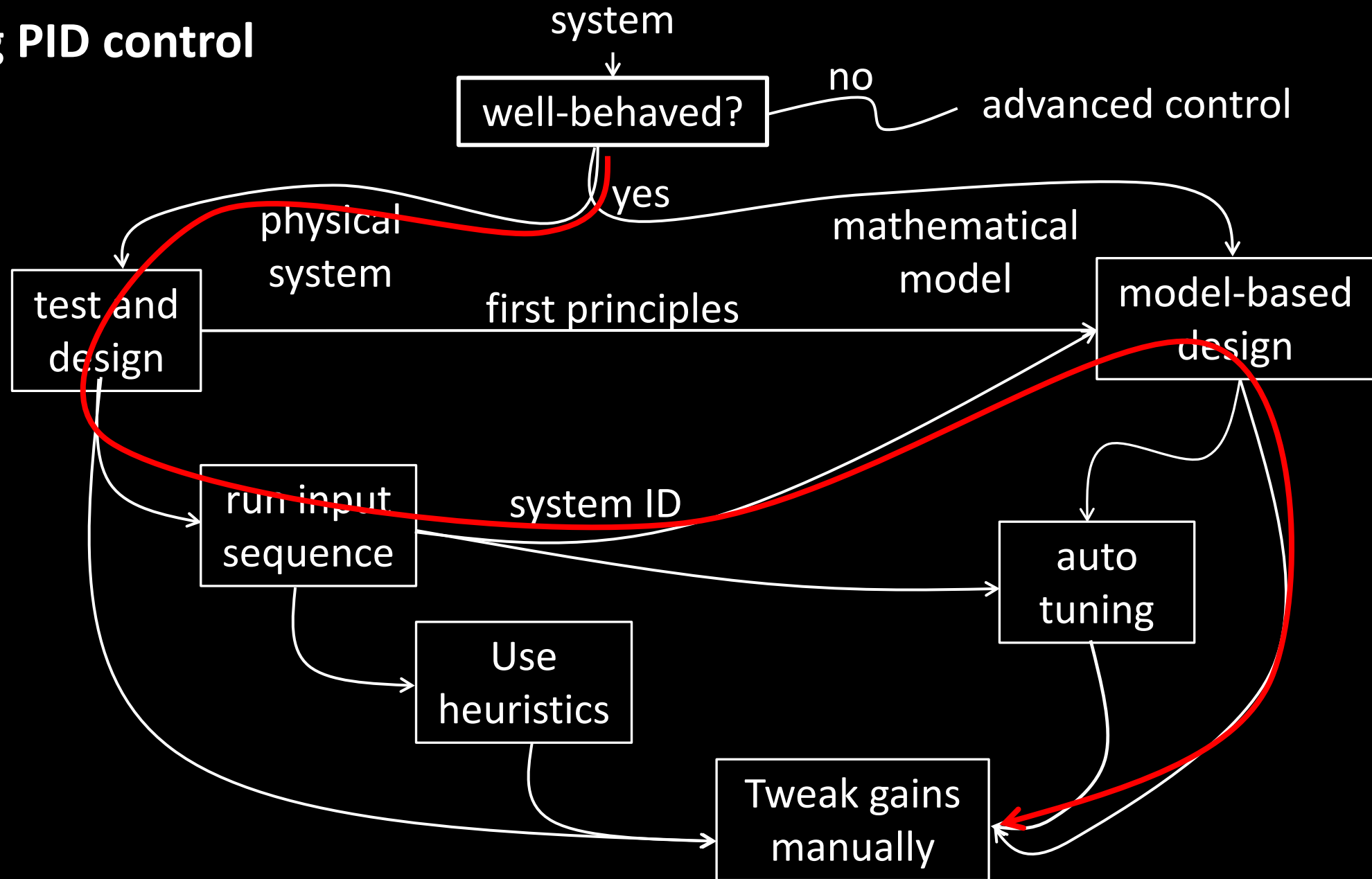


# PID control

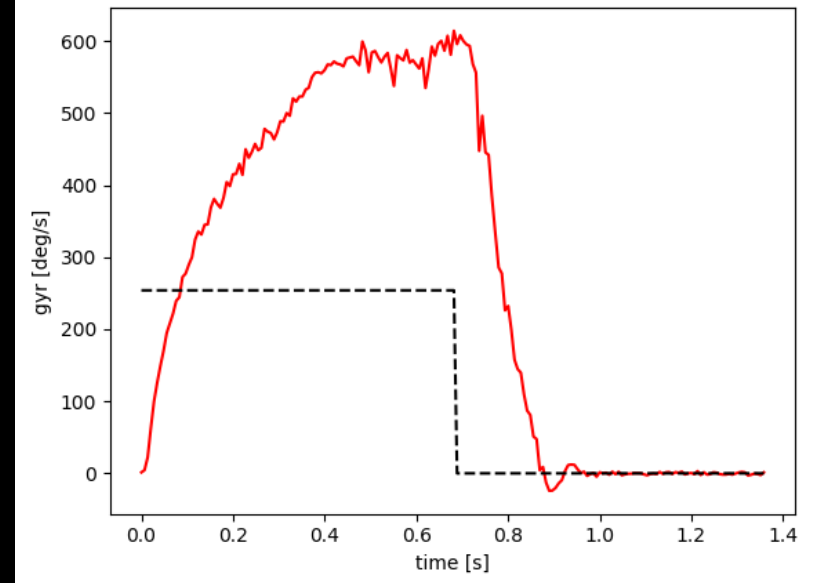
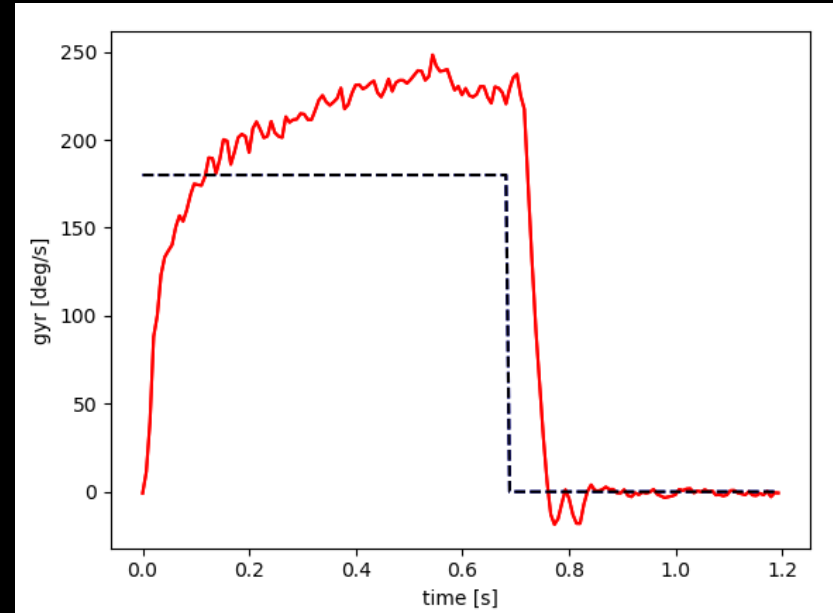
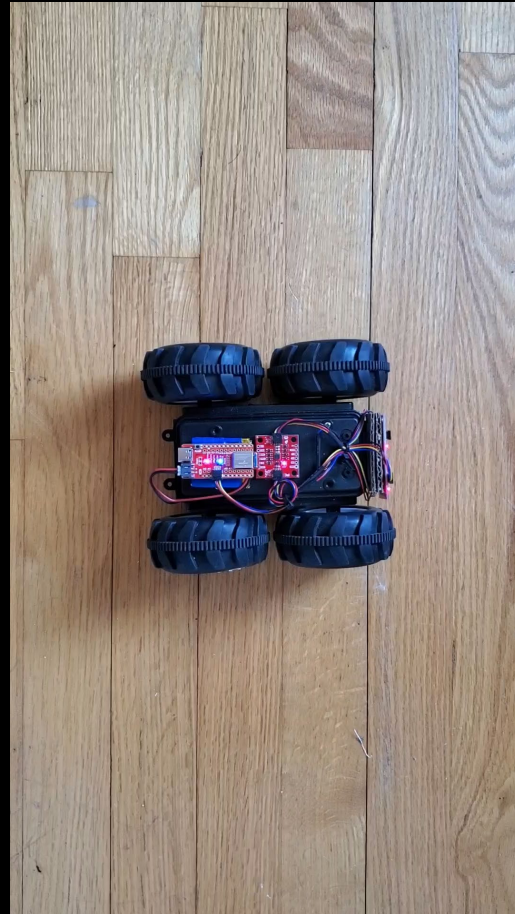
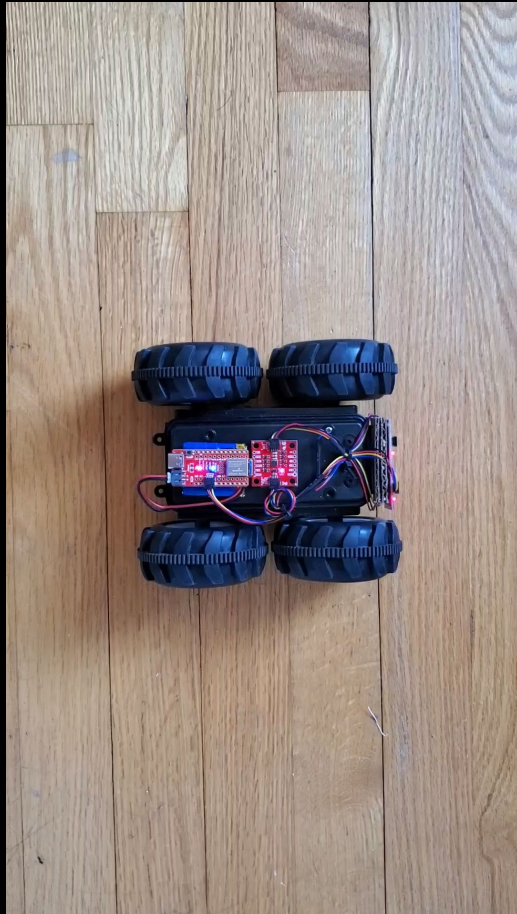
- **Heuristic procedure #1:**
  - Set  $K_p$  to small value,  $K_D$  and  $K_I$  to 0
  - Increase  $K_D$  until oscillation, then decrease by factor of 2-4
  - Increase  $K_P$  until oscillation or overshoot, decrease by factor of 2-4
  - Increase  $K_I$  until oscillation or overshoot
  - Iterate
- **Heuristic procedure #2:**
  - Set  $K_D$  and  $K_I$  to 0
  - Increase  $K_P$  until oscillation, then decrease by factor of 2-4
  - Increase  $K_I$  until loss of stability, then back off
  - Increase  $K_D$  to increase performance in response to disturbance
  - Iterate



# Tuning PID control



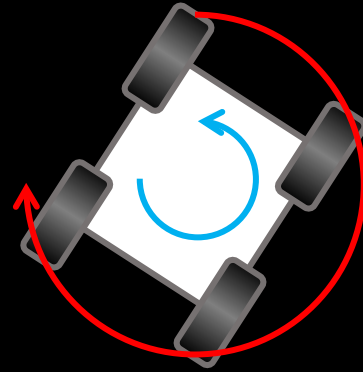
# Tuning PID control



# Tuning PID control

- Equations of motion

- $x = \begin{bmatrix} \theta \\ \dot{\theta} \end{bmatrix}$



- <https://tinyurl.com/y67glgzk>

$$F = ma$$

$$\tau = I\alpha$$

$$\tau = I\ddot{\theta}$$

$$u - \dot{\theta}c = I\ddot{\theta}$$

$$\ddot{\theta} = \frac{-\dot{\theta}c}{I} + \frac{1}{I}u$$

$$\begin{bmatrix} \dot{\theta} \\ \ddot{\theta} \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 0 & \frac{-c}{I} \end{bmatrix} \begin{bmatrix} \theta \\ \dot{\theta} \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{I} \end{bmatrix} u$$

